



Design Example Report

Title	<i>5.8 W TRIAC Dimmable, High Efficient, High Power Factor Corrected (>0.9), Non-Isolated Buck-Boost LED Driver Using LinkSwitch™ - PL LNK460KG</i>
Specification	195 VAC – 265 VAC Input; 145 V _{TYP} , 40 mA Output
Application	Dimmable GU10 LED Driver
Author	Applications Engineering Department
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Revision	1.0

Summary and Features

- Single-stage power factor correction combined with constant current (CC) output
- Low cost, low component count, highly efficient, and fits in GU10 LED lamp
- Fast start-up time (<300 ms) – no perceptible delay
- Integrated protection and reliability features
 - Output short-circuit protected with auto-recovery
 - Auto-recovering thermal shutdown with large hysteresis protects both components and PCB
 - No damage during brown-out conditions
- PF >0.9 at 230 VAC
- Compatible with most leading edge and trail edge TRIAC dimmers

PATENT INFORMATION

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Table of Contents

1	Introduction.....	4
2	Power Supply Specification	6
3	Schematic.....	7
4	Circuit Description	8
4.1	Input EMI Filtering	8
4.2	Power and Feedback Circuits.....	8
4.3	TRIAC Dimming Control Compatibility	8
5	PCB Layout	10
6	Bill of Materials	11
7	Inductor Specification	12
7.1	Electrical Diagram	12
7.2	Electrical Specifications.....	12
7.3	Materials.....	12
7.4	Inductor Build Diagram.....	13
7.5	Inductor Instructions	13
8	Inductor Design Spreadsheet	14
9	Performance Data	16
9.1	Efficiency	16
9.2	Line and Load Regulation.....	17
9.3	Power Factor	18
9.4	A-THD	19
9.5	Harmonics	20
9.5.1	140 V LED Load	20
9.5.2	145 V LED Load	21
9.5.3	150 V LED Load	22
9.6	Test Data.....	23
9.6.1	Test Data, 140 V LED Load	23
9.6.2	Test Data, 145 V LED Load	23
9.6.3	Test Data, 150 V LED Load	23
9.6.4	230 VAC 50 Hz, 140 V LED Load Harmonics Data.....	24
9.6.5	230 VAC 50 Hz, 145 V LED Load Harmonics Data.....	25
9.6.6	230 VAC 50 Hz, 150 V LED Load Harmonics Data.....	26
10	Thermal Performance	27
11	Waveforms.....	28
11.1	Input Voltage and Input Current Waveforms	28
11.2	Output Current and Output Voltage at Normal Operation.....	29
11.3	Output Current/Voltage Rise and Fall.....	30
11.4	Input Voltage and Output Current Waveform at Start-up.....	31
11.5	Drain Voltage and Current at Normal Operation.....	32
11.6	Start-up Drain Voltage and Current	33
11.7	Drain Current and Drain Voltage During Output Short Condition	33
11.8	Freewheeling Diode Voltage and Output Current Waveform.....	34
11.9	Input Voltage and Output Current Waveform with a Leading Edge Dimmer.....	35
11.10	Input Voltage and Output Current Waveform with a Trailing Edge Dimmer ...	36



11.11 Voltage Waveforms Across Damping Resistor and Bleeder Resistor37
11.12 Output Voltage Waveform in Open Load Condition37
12 Conducted EMI38
13 Revision History39

Important Note: Although this board is designed to satisfy safety requirements, the engineering prototype has not been agency approved. In addition, this design does not provide galvanic isolation of the output from the AC input. Therefore, all testing should be performed using an isolation transformer to provide the AC input to the prototype board.



1 Introduction

The document describes a dimmable non-isolated, high efficiency, high power factor (PF) LED driver designed to drive a nominal LED string voltage of 145 V at 40 mA from an input voltage range of 195 VAC to 265 VAC. The LED driver utilizes the LNK460KG from the LinkSwitch-PL family of ICs.

The topology used is a single-stage non-isolated buck-boost that fits in the GU10 lamp form factor and meets the high efficiency requirements for this design. LinkSwitch-PL based designs provide high power factor (>0.9), meeting international harmonic requirements.

This document contains the LED driver specification, schematic, PCB details, bill of materials, transformer documentation and typical performance characteristics.

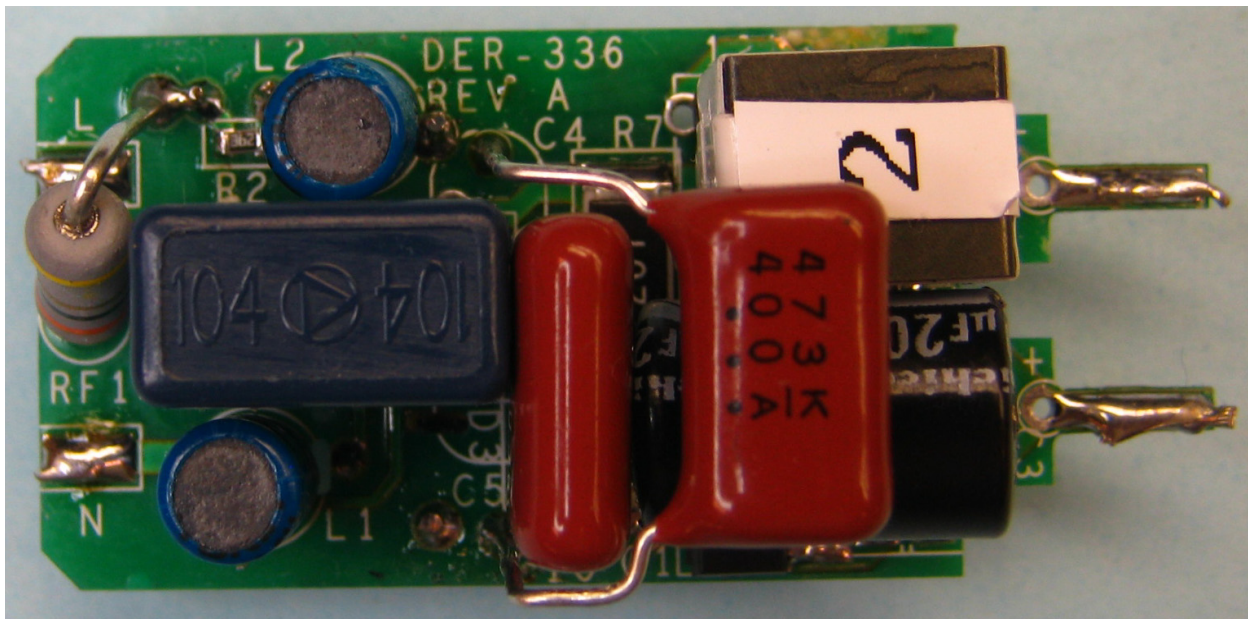


Figure 1 – Populated Circuit Board, Top View.



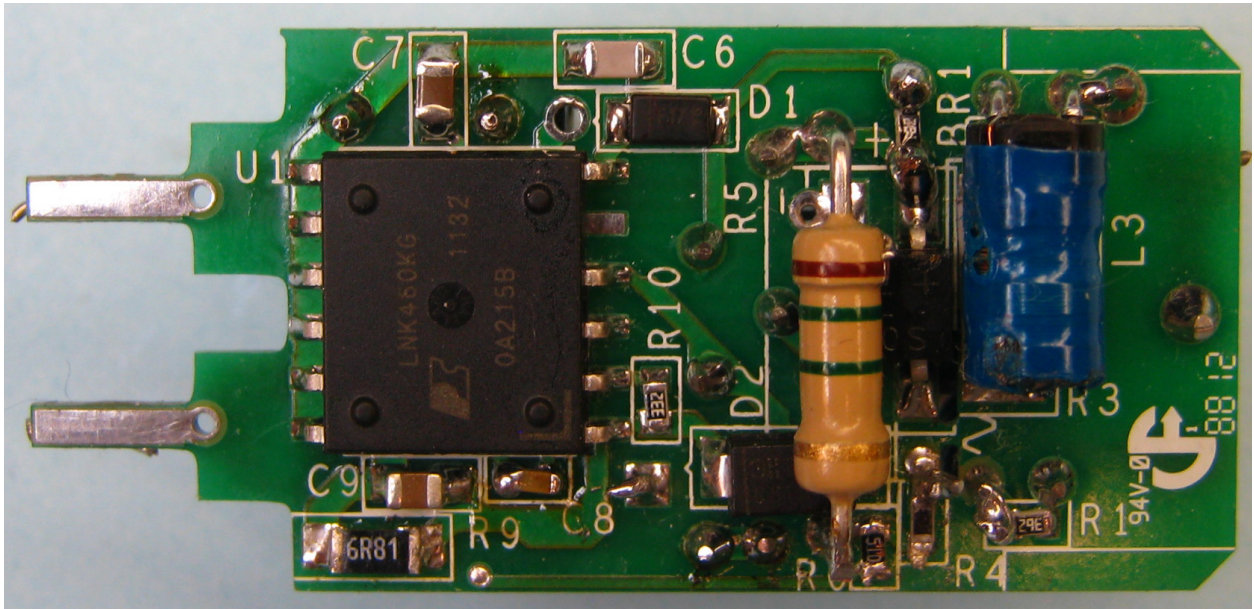


Figure 2 – Populated Circuit Board, Bottom View.

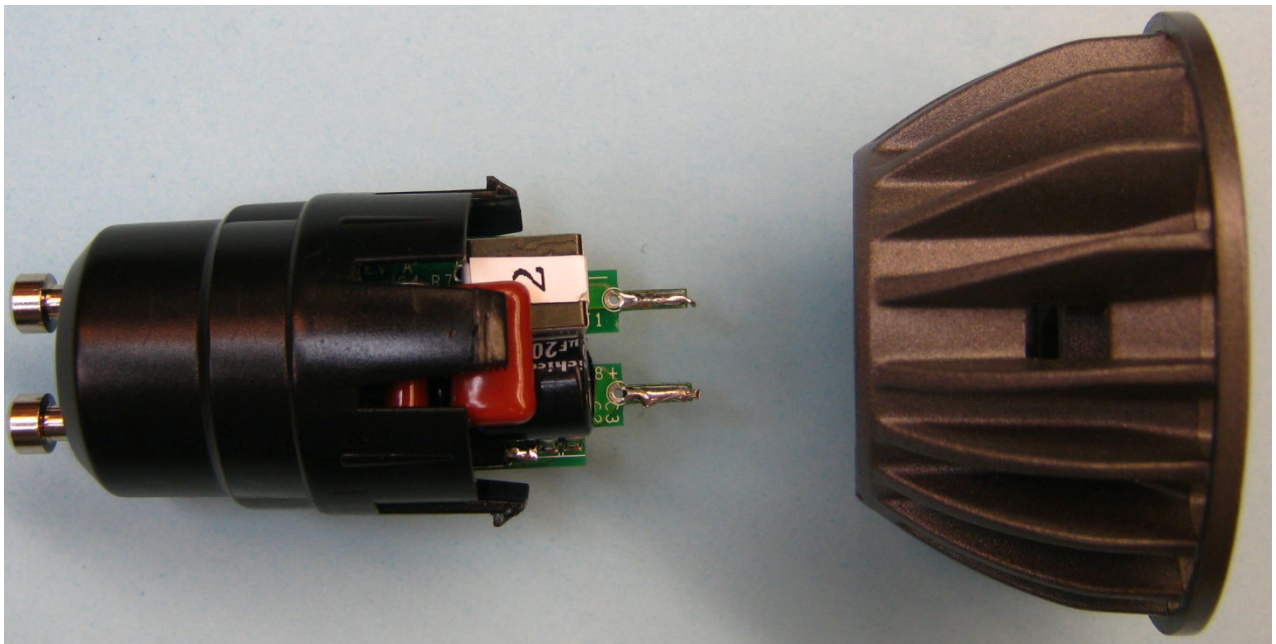


Figure 3 – Fitted into a GU10 Case.



2 Power Supply Specification

The table below represents the minimum acceptable performance of the design. Actual performance is listed in the results section.

Description	Symbol	Min	Typ	Max	Units	Comment
Input Voltage	V_{IN}	195	235	265	VAC	2 Wire – no P.E.
Frequency	f_{LINE}		50		Hz	
Output Output Voltage	V_{OUT}		145		V	
Output Current	I_{OUT}		40		mA	
Total Output Power Continuous Output Power	P_{OUT}		5.8		W	
Efficiency Full Load	η		80		%	Measured at P_{OUT} 25 °C
Environmental Conducted EMI			CISPR 15B / EN55015B			NOT TESTED
Safety			Non-Isolated			
Ring Wave (100 kHz) Differential Mode (L1-L2)						
Differential Surge (1.2 / 50 μ s)						
Power Factor		0.9				At 230 VAC, 50 Hz
Harmonic Currents			EN 61000-3-2 Class D (C)			Class C specifies Class D Limits when $P_{IN} < 25$ W
Ambient Temperature	T_{AMB}		40		°C	Free convection, sea level



3 Schematic

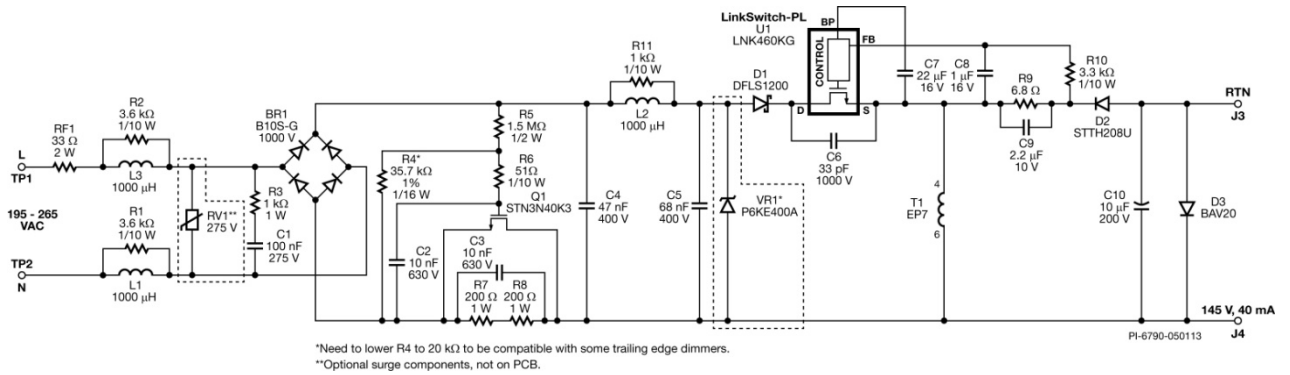


Figure 4 – Schematic.



4 Circuit Description

The LinkSwitch-PL (U1) is a highly integrated primary-side controller intended for use in LED driver applications. The LinkSwitch-PL provides high power factor while regulating the output current in a single conversion stage. The design also compensates for the output voltage variations typically encountered in LED driver applications. All of the control circuitry responsible for these functions plus the high-voltage power MOSFET is incorporated into the IC.

4.1 Input EMI Filtering

Inductors L1-L3 and C4-C5 filters the input switching current presented by the buck-boost converter to the line. Resistor R1, R2 and R3 across L1, L2 and L3 damp any resonances between the input inductors, capacitors and the AC line impedance which would ordinarily show up as increased conducted EMI.

Bridge rectifier BR1 rectifies the AC line voltage with capacitor C4 providing a low impedance path (decoupling) for the primary switching current. A low capacitance value (sum of C4 and C5) is necessary to maintain a power factor of greater than 0.9.

4.2 Power and Feedback Circuits

The circuit is configured as a buck-boost converter with the SOURCE (S) pin of U1 connected to the cathode of the freewheeling diode D2 via a current sense resistor. The current sense resistor R9 is used to sense the diode current of the buck-boost converter. Its value is adjusted to center the output current at 40 mA at nominal input voltage. Capacitor C8 and R10 act as low pass filter to average the diode current which is used as feedback signal which is proportional to output current. Capacitor C9 acts as bypass for the high frequency passing through R9, improving overall efficiency.

The DRAIN (D) pin connected to the positive side of the DC rectified input thru D1. Diode D1 is used to prevent reverse current from flowing through U1. An EP7 inductor core size was optimized to fit in the GU10 case.

Capacitor C7 provides local decoupling for the BYPASS (BP) pin of U1 which is the supply pin for the internal controller. During start-up, C7 is charged to ~6 V from an internal high-voltage current source connected to the DRAIN pin.

Diode D3 provides disconnected load protection, failing as a short-circuit when the load is disconnected.

4.3 TRIAC Dimming Control Compatibility

The active damper consists of components R4, R5, R6, C2, Q1 in conjunction with R7 and R8. This circuit limits the inrush current that flows to charge C4 when the TRIAC turns on by placing R7 and R8 in series for approximately 2.5 ms each half line cycle. This keeps the power dissipation on R7 and R8 low and allows a larger value during



current limiting. Resistor divider R4 and R5 determine turn on threshold of the input voltage.

The passive bleeder circuit is comprised of X capacitor C1 and R3. This keeps the input current above the TRIAC holding current while the input current corresponding to the driver increases during each AC half-cycle preventing the TRIAC oscillating on and off at the start of each conduction period.



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5 PCB Layout

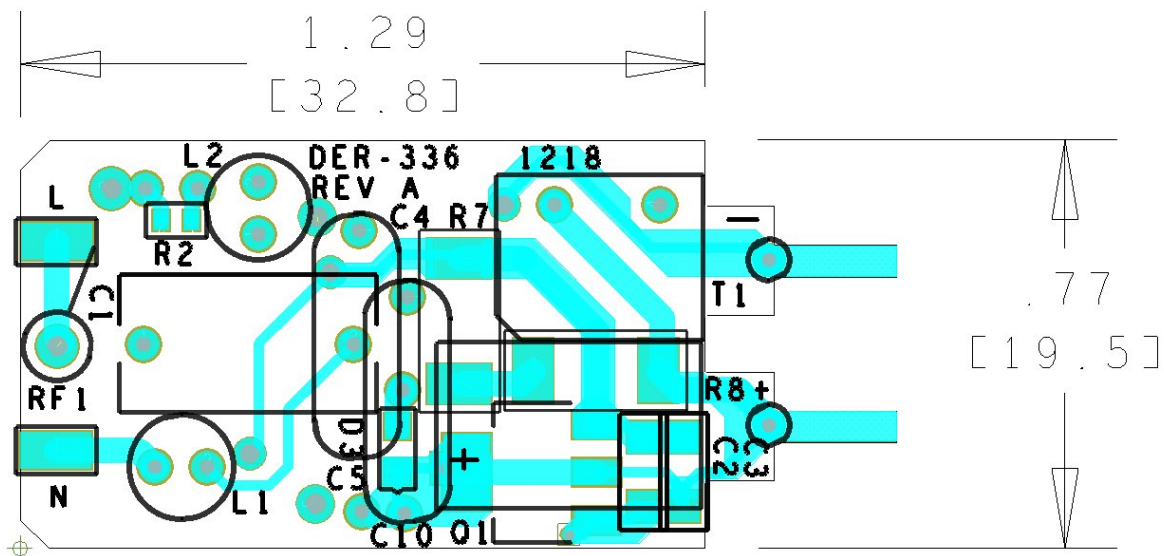


Figure 5 – Top Side.

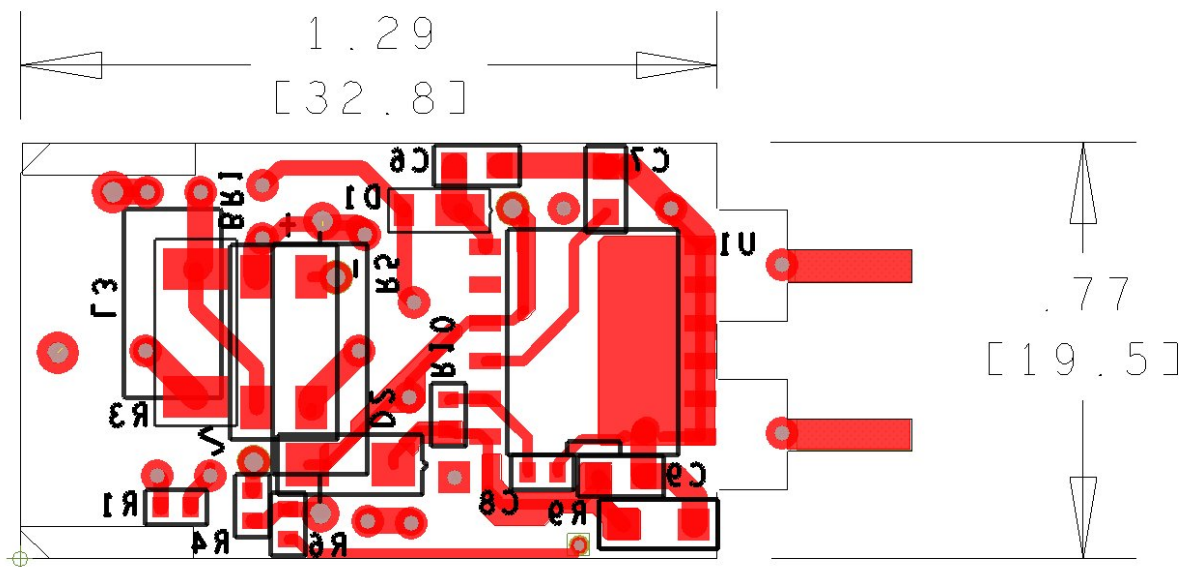


Figure 6 – Bottom Side.



6 Bill of Materials

Item	Qty	Ref Des	Description	Mfg Part Number	Mfg
1	1	BR1	1000 V, 0.8 A, Bridge Rectifier, SMD, MBS-1, 4-SOIC	B10S-G	Comchip Tech
2	1	C1	100 nF, 275 VAC, Film, X2	LE104-M	OKAY
3	2	C2 C3	10 nF, 630 V, Ceramic, X7R, 1206	C1206C103KBRACU	Kemet
4	1	C4	47 nF, 400 V, Film	ECQ-E4473KF	Panasonic
5	1	C5	68 nF, 400 V, Film	ECQ-E4683KF	Panasonic
6	1	C6	33 pF, 1000 V, Ceramic, COG, 0805	0805AA330KAT1A	AVX
7	1	C7	22 μ F, 16 V, Ceramic, X7R, 0805	C2012X5R1C226K	TDK
8	1	C8	1 μ F, 16 V, Ceramic, X5R, 0603	GRM188R61C105KA93D	Murata
9	1	C9	2.2 μ F, 10 V, Ceramic, X7R, 0805	GRM21BR71A225MA01L	Murata
10	1	C10	10 μ F, 200 V, Electrolytic, (8 x 11)	SMQ200VB10RM8X11LL	Nippon Chemi-Con
11	1	D1	200 V, 1 A, Diode Schottky, PWRD1123	DFLS1200	Diodes, Inc.
12	1	D2	800 V, 2 A, Ultrafast Recovery, 75 ns, DO-214AA	STTH208U	ST Micro
13	1	D3	200 V, 200 mA, Fast Switching, 50 ns, DO-35	BAV20	Vishay
14	3	L1 L2 L3	1000 μ H, 0.21 A, 5.5 x 10.5 mm	SBC1-102-211	Token
15	1	Q1	400 V, 1.8 A, N-Channel, SOT 223	STN3N40K3	ST Micro
16	2	R1 R2	3.6 k Ω , 5%, 1/10 W, Thick Film, 0603	ERJ-3GEYJ362V	Panasonic
17	1	R3	1 k Ω , 5%, 1 W, Thick Film, 2512	ERJ-1TYJ102U	Panasonic
18	1	R4	35.7 k Ω , 1%, 1/16 W, Thick Film, 0603	ERJ-3EKF3572V	Panasonic
19	1	R5	1.5 M Ω , 5%, 1/2 W, Carbon Film	CFR-50JB-1M5	Yageo
20	1	R6	51 Ω , 5%, 1/10 W, Thick Film, 0603	ERJ-3GEYJ510V	Panasonic
21	2	R7 R8	200 Ω , 5%, 1 W, Thick Film, 2512	ERJ-1TYJ201U	Panasonic
22	1	R9	6.8 Ω , 5%, 1/4 W, Thick Film, 1206	ERJ-8GEYJ6R8V	Panasonic
23	1	R10	3.3 k Ω , 5%, 1/10 W, Thick Film, 0603	ERJ-3GEYJ332V	Panasonic
24	1	R11	1 k Ω , 5%, 1/10 W, Thick Film, 0603	ERJ-3GEYJ102V	Panasonic
25	1	RF1	33 Ω , 2 W, Fusible/Flame Proof Wire Wound	ULW2-33RJA25	TT Electronics-Welwyn
26	1	T1	Bobbin, EP7, 6 pins	CSH-EP7-1S-6P-E	Phillips
27	1	U1	LinkSwitch-PL, eSOP-12B	LNK460KG	Power Integrations



7 Inductor Specification

7.1 Electrical Diagram

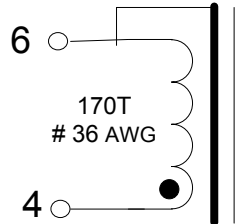


Figure 7 – Inductor Electrical Diagram.

7.2 Electrical Specifications

Primary Inductance	Pins 4-6, all other windings open, measured at 66 kHz, 0.4 V _{RMS} .	330 μ H \pm 5%
Resonant Frequency	Pins 4-6, all other windings open.	1 MHz (Min.)

7.3 Materials

Item	Description
[1]	Core: EP7.
[2]	Bobbin: B-EP7-V-6pins-(3/3).
[3]	Tape, Polyester film, 3M 1350F-1 or equivalent, 6.4 mm wide.
[4]	Wire: Magnet, #36 AWG, solderable double coated.
[5]	Copper Tape: 2 mil thick.



7.4 Inductor Build Diagram

Pins Side

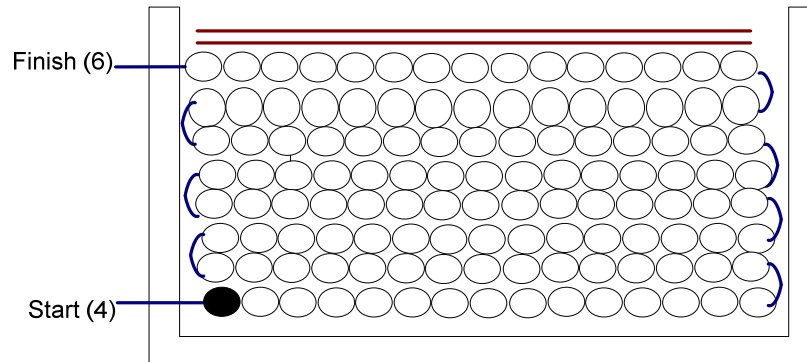


Figure 8 – Inductor Build Diagram.

7.5 Inductor Instructions

Bobbin Preparation	Place the bobbin item [2] on the mandrel such that pin side on the left side. Winding direction is the clockwise direction.
WDG	Starting at pin 4, wind 170 turns of wire item [4] in 8 layers. Finish at pin 6.
Core Assembly	Grind core to get 0.33 mH inductance. Assemble and secure core halves
Flux Band	Construct a flux band by wrapping a single shorted turn of item [5] around the output side of windings and core halves with tight tension. Make an electrical connection to pin (6) using wire. Add 3 layers of tape, item (3) for insulation
Varnish	Dip varnish uniformly.

8 Inductor Design Spreadsheet

ACDC_LinkSwitch-PL-Buck-Boost_121211; Rev.1.0; Copyright Power Integrations 2011	INPUT	INFO	OUTPUT	UNIT	ACDC_LinkSwitch-PL-Buck-Boost_121211; LinkSwitch-PL Buck-Boost Transformer Design Spreadsheet
ENTER APPLICATION VARIABLES					
VACMIN	195		195	V	Minimum AC input voltage
VACNOM	230		230	V	Nominal AC input voltage
VACMAX	265		265	V	Maximum AC input voltage
FL			50	Hz	Minimum line frequency
VO_MIN			136.8	V	Minimum output voltage tolerance
VO_NOM	144.00		144.00	V	Nominal Output Voltage
VO_MAX			151.20	V	Maximum output voltage tolerance
IO	0.040		0.040	A	Average output current specification
n	0.80		0.800	%/100	Total power supply efficiency
Z			0.5		Loss allocation factor
Enclosure	Retrofit Lamp		Retrofit Lamp		Enclosure selections determine thermal conditions and maximum power. Enter "Retrofit Lamp" or "Open frame"
PO			5.76	W	Total output power
VD			0.4	V	Output diode forward voltage drop
LinkSwitch-PL DESIGN VARIABLES					
Device	LNK460		LNK460		Chosen LinkSwitch-PL Device
TON			1.34	us	Expected on-time of MOSFET at low line and PO
FSW			45.4	kHz	Expected switching frequency at low line and PO
Duty Cycle			6.1	%	Expected operating duty cycle at low line and PO
VDRAIN			546	V	Estimated worst case drain voltage at VACMAX and VO_MAX
IRMS			0.133	A	Nominal RMS current through the switch
IPK			1.617	A	Worst Case Peak current
ILIM_MIN			1.637	A	Minimum device current limit
KDP			6.86		Ratio between off-time of switch and reset time of core at VACNOM
LinkSwitch-PL EXTERNAL COMPONENT CALCULATIONS					
RSENSE			7.250	Ohms	Output current sense resistor
Standard RSENSE			7.32	Ohms	Closest 1% value for RSENSE
PSENSE			11.6	mW	Power dissipated by RSENSE
ENTER INDUCTOR CORE/CONSTRUCTION VARIABLES					
Core Type	EP7		EE10		Core Type
AE	10.70		10.70	mm ²	Core Effective Cross Sectional Area
LE	15.50		15.50	mm	Core Effective Path Length
AL	1000		1000	nH/T ²	Ungapped Core Effective Inductance
BW	3.50		3.5	mm	Bobbin Physical Winding Width
L	8		8		Number of winding layers
TRANSFORMER PRIMARY DESIGN PARAMETERS					
LP			330.0	uH	Primary Inductance
LP Tolerance			10	%	Tolerance of Primary Inductance
N	170.00		170	Turns	Number of Turns
ALG			11	nH/T ²	Gapped Core Effective Inductance
BM			2934	Gauss	Operating Flux Density
BAC			1467	Gauss	Worst case AC Flux Density for Core Loss Curves (0.5 X Peak to Peak)



BP		Warning	4157	Gauss	!!! Reduce peak flux density (BP < 3600 G) by increasing NP, selecting a bigger core or decreasing KDP
LG			1.178	mm	Gap Length (Lg > 0.1 mm)
BWE			28	mm	Effective Bobbin Width
L_IRMS			0.258	A	RMS Current through the inductor
OD			0.16	mm	Maximum Primary Wire Diameter including insulation
INS			0.04	mm	Estimated Total Insulation Thickness (= 2 * film thickness)
DIA			0.13	mm	Bare conductor diameter
AWG			36	AWG	Primary Wire Gauge (Rounded to next smaller standard AWG value)
CM			25	Cmils	Bare conductor effective area in circular mils
CMA		Warning	98	Cmils/Amp	!!! INCREASE CMA > 200 (increase L (primary layers) or choose a larger core)
Current Density (J)		Warning	20.37	A/mm ²	!!! Current density is above recommended value of 9.75 A/mm ² . Use larger wire diameter (OD), increase L or increase core size to decrease current density.
Output Parameters					
IO			0.040	A	Expected Output Current
PIVD			737.0	V	Peak Inverse Voltage at VO_MAX on output diode

Notes:

1. The peak flux density warning (BP) can be ignored for this design. The spreadsheet BP calculation assumes that the LNK460KG will operate at $I_{LIM(MAX)}$ during start-up. In practice, due to the internal soft-start function this current level is not reached and therefore no core saturation occurs. This was confirmed in both Figures 38 and 39 for normal start-up and Figures 41 and 43 for start-up with a shorted output (fault condition). In all cases, the peak drain current is below the absolute maximum data sheet specification
2. CMA <200 Cmils/Amp and J>9.75 A/mm² are acceptable in this design with low inductor temperature.



9 Performance Data

All measurements performed at room temperature using an LED load. The following data were measured using 3 sets of loads to represent the load range of 140 V to 150 V (output voltage). Refer to the table on Section 9.6 for the complete set of test data values.

9.1 Efficiency

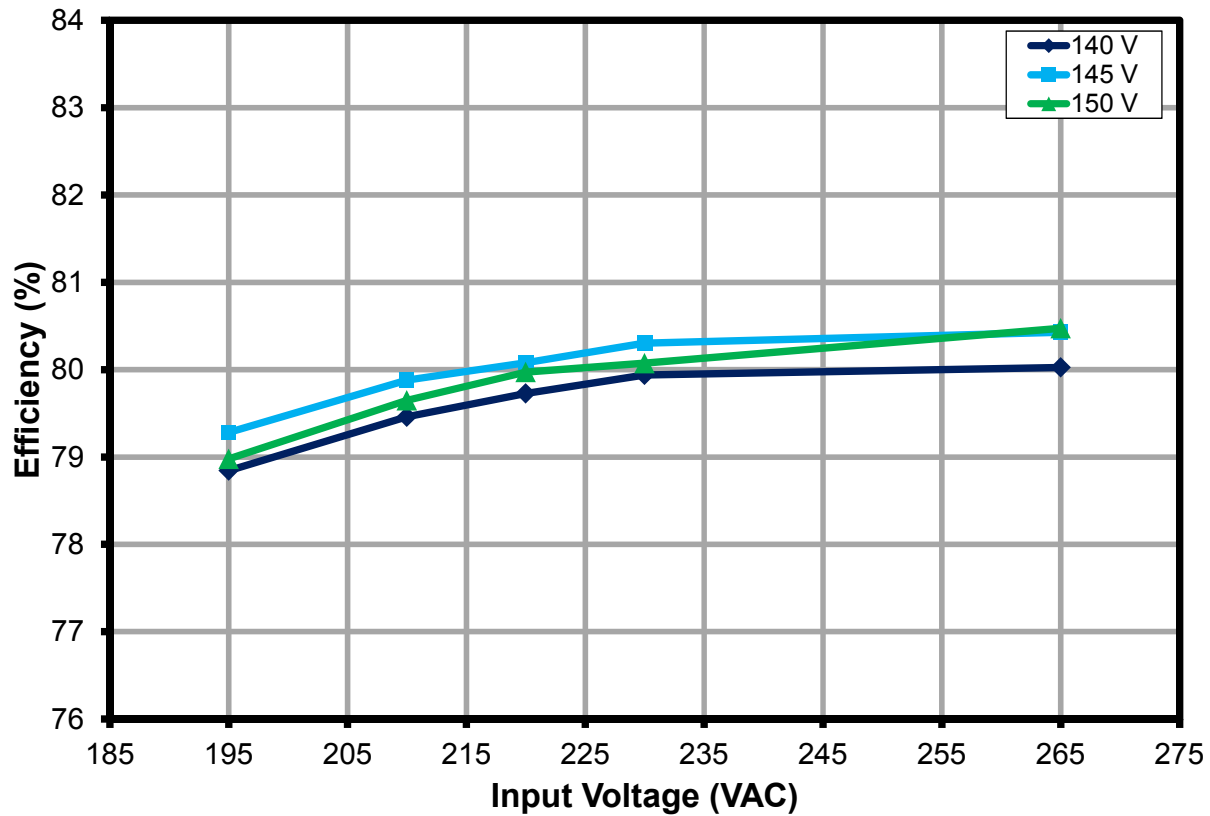


Figure 9 – Efficiency vs. Line and Load.



9.2 Line and Load Regulation

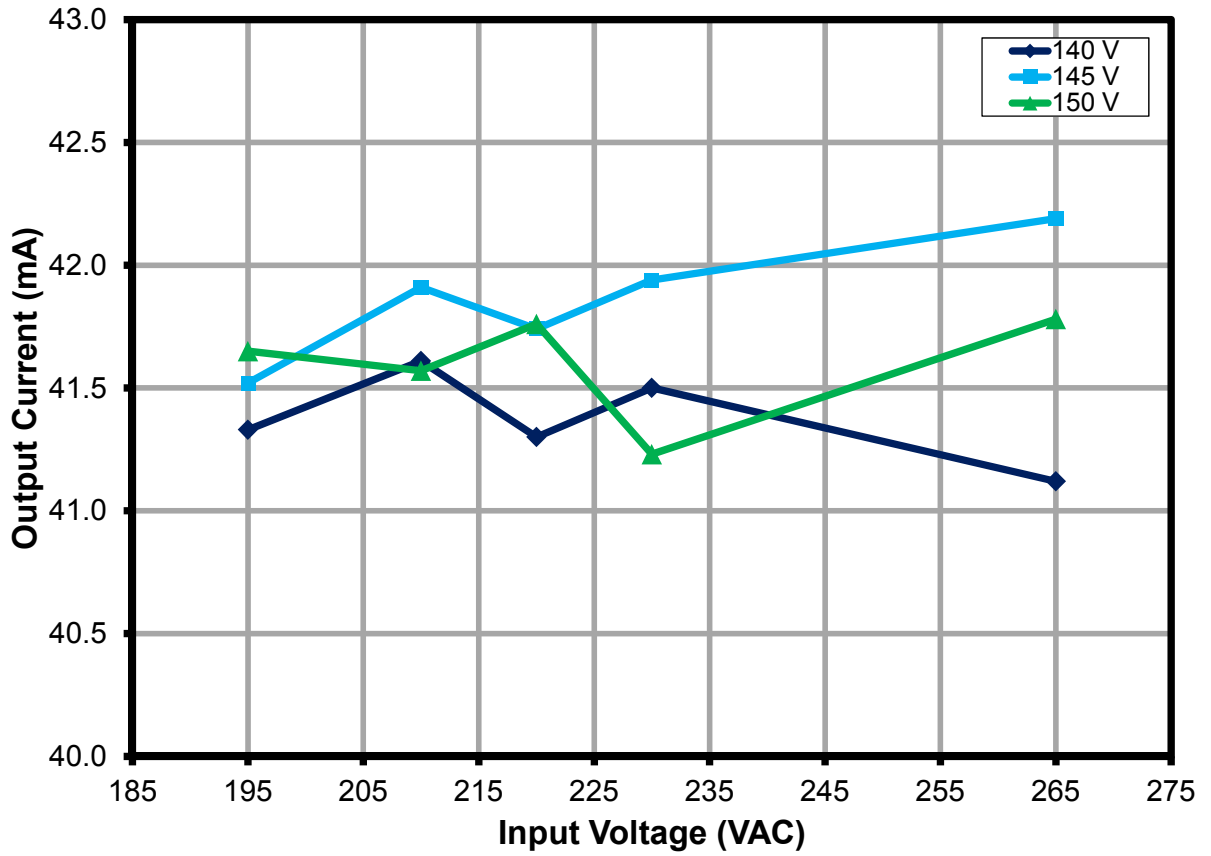


Figure 10 – Regulation vs. Line and Load.



9.3 Power Factor

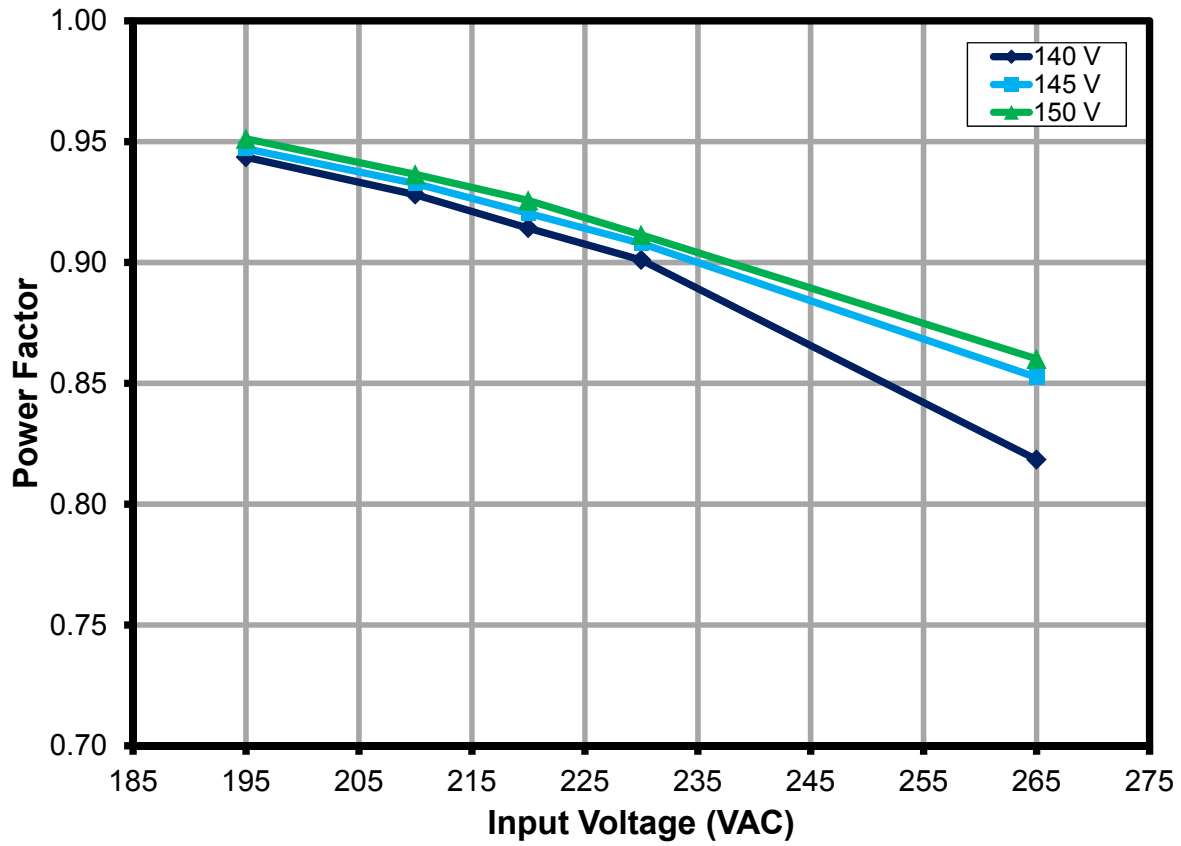


Figure 11 – Power Factor vs. Line and Load.



9.4 A-THD

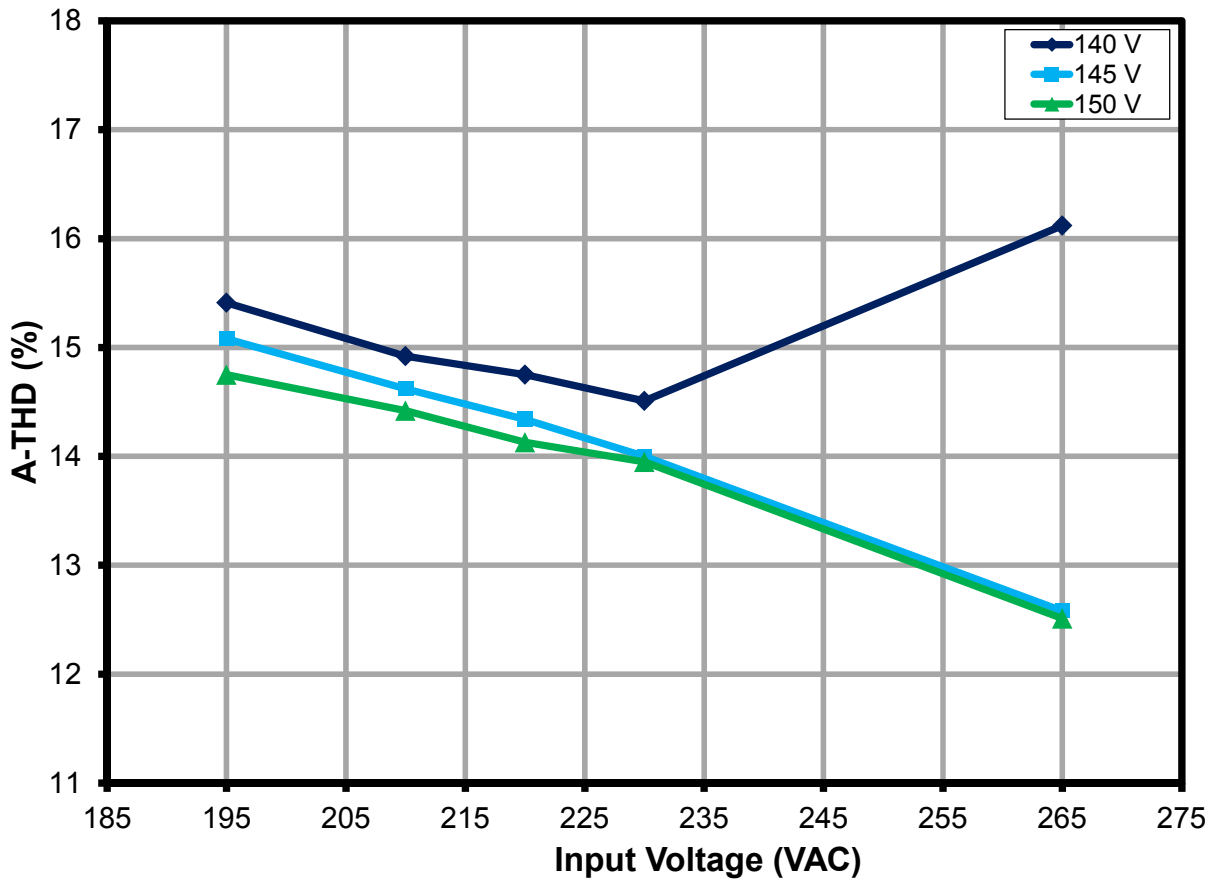


Figure 12 – A-THD vs. Line and Load.



9.5 Harmonics

The design met the limits for Class C equipment for an active input power of <25 W. In this case IEC61000-3-2 specifies that harmonic currents shall not exceed the limits of Class D equipment¹. Therefore the limits shown in the charts below are Class D limits which must not be exceeded to meet Class C compliance.

9.5.1 140 V LED Load

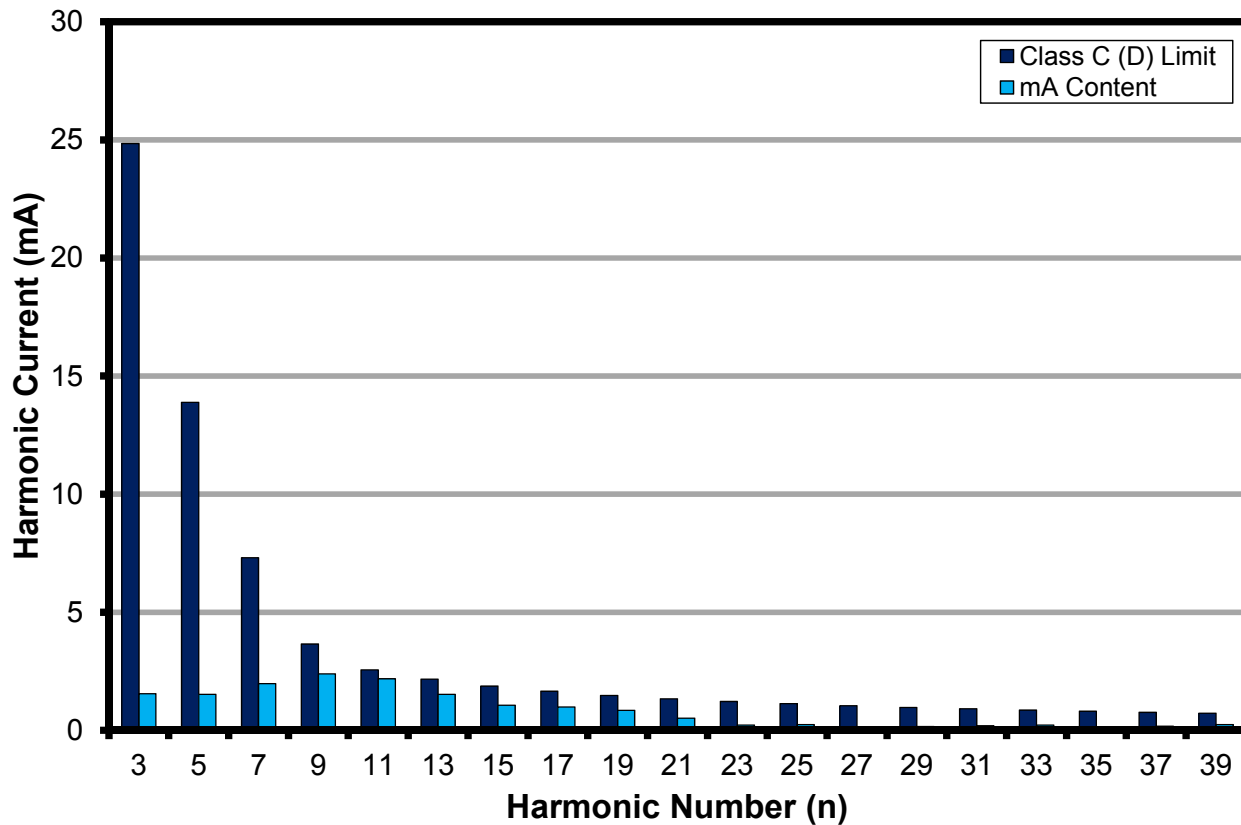


Figure 13 – 140 V LED Load Input Current Harmonics at 230 VAC, 50 Hz.

¹ IEC6000-3-2 Section 7.3, table 2, column 2.



9.5.2 145 V LED Load

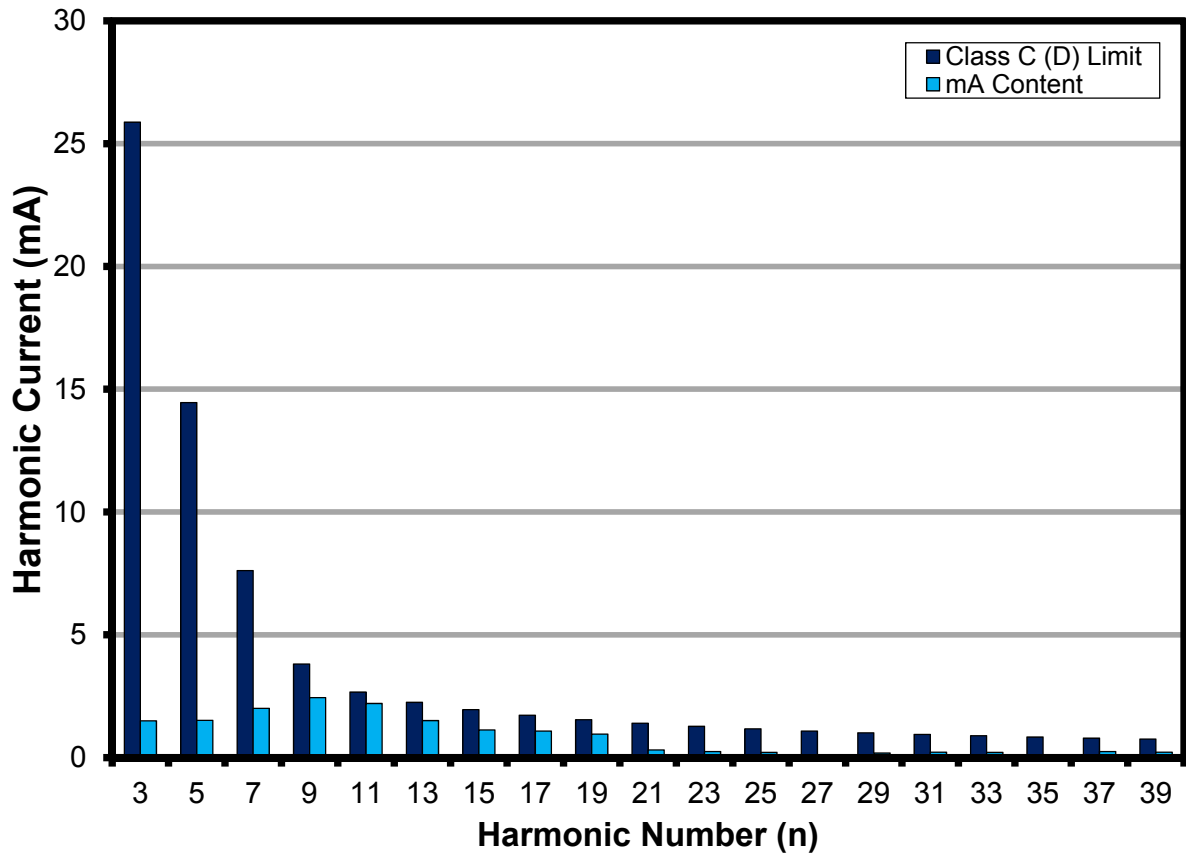


Figure 14 – 145 V LED Load Input Current Harmonics at 230 VAC, 50 Hz.



9.5.3 150 V LED Load

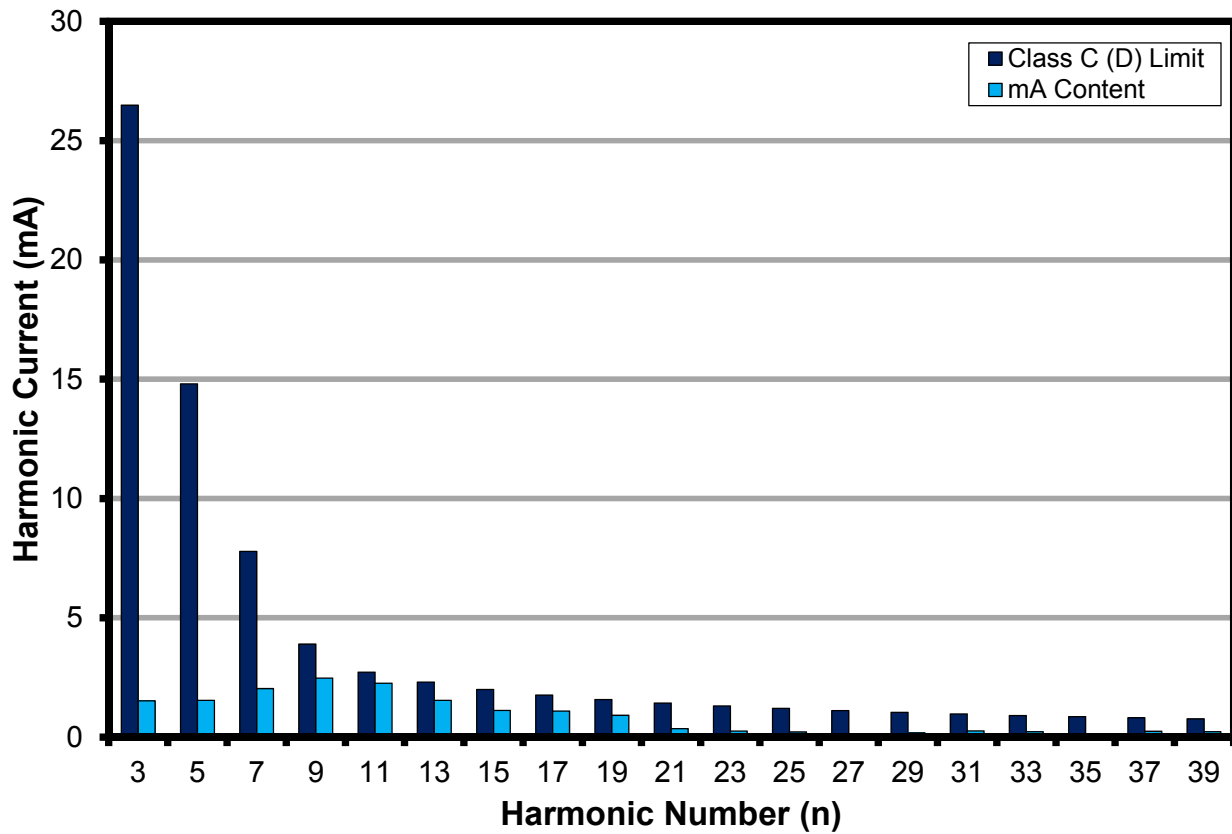


Figure 15 – 150 V LED Load Input Current Harmonics at 230 VAC, 50 Hz.



9.6 Test Data

All measurements were taken with the board at open frame, 25 °C ambient.

9.6.1 Test Data, 140 V LED Load

Input		Input Measurement					Load Measurement			Calculation		
VAC (V _{RMS})	Freq (Hz)	V _{IN} (V _{RMS})	I _{IN} (mA _{RMS})	P _{IN} (W)	PF	%ATHD	V _{OUT} (V _{DC})	I _{OUT} (mA _{DC})	P _{OUT} (W)	P _{CAL} (W)	Efficiency (%)	Loss (W)
195	50	194.96	40.13	7.383	0.944	15.41	139.3000	41.330	5.821	5.76	78.84	1.56
210	50	209.90	37.87	7.376	0.928	14.92	139.3000	41.610	5.861	5.80	79.46	1.52
220	50	219.94	36.26	7.291	0.914	14.75	139.3000	41.300	5.813	5.75	79.73	1.48
230	50	229.98	35.27	7.308	0.901	14.51	139.3000	41.500	5.842	5.78	79.94	1.47
265	50	265.00	33.34	7.229	0.818	16.12	139.3000	41.120	5.785	5.73	80.02	1.44

9.6.2 Test Data, 145 V LED Load

Input		Input Measurement					Load Measurement			Calculation		
VAC (V _{RMS})	Freq (Hz)	V _{IN} (V _{RMS})	I _{IN} (mA _{RMS})	P _{IN} (W)	PF	%ATHD	V _{OUT} (V _{DC})	I _{OUT} (mA _{DC})	P _{OUT} (W)	P _{CAL} (W)	Efficiency (%)	Loss (W)
195	50	194.96	41.36	7.636	0.947	15.08	144.2000	41.520	6.054	5.99	79.28	1.58
210	50	209.89	39.07	7.649	0.933	14.62	144.3000	41.910	6.110	6.05	79.88	1.54
220	50	219.94	37.52	7.595	0.920	14.34	144.2000	41.740	6.082	6.02	80.08	1.51
230	50	229.98	36.45	7.610	0.908	14	144.2000	41.940	6.111	6.05	80.30	1.50
265	50	264.99	33.84	7.644	0.853	12.58	144.3000	42.190	6.148	6.09	80.43	1.50

9.6.3 Test Data, 150 V LED Load

Input		Input Measurement					Load Measurement			Calculation		
VAC (V _{RMS})	Freq (Hz)	V _{IN} (V _{RMS})	I _{IN} (mA _{RMS})	P _{IN} (W)	PF	%ATHD	V _{OUT} (V _{DC})	I _{OUT} (mA _{DC})	P _{OUT} (W)	P _{CAL} (W)	Efficiency (%)	Loss (W)
195	50	194.95	43.07	7.987	0.951	14.75	149.8000	41.650	6.308	6.24	78.98	1.68
210	50	209.88	40.20	7.901	0.937	14.42	149.8000	41.570	6.293	6.23	79.65	1.61
220	50	219.93	38.83	7.904	0.926	14.13	149.8000	41.760	6.321	6.26	79.97	1.58
230	50	229.97	37.16	7.789	0.911	13.95	149.8000	41.230	6.237	6.18	80.07	1.55
265	50	264.99	34.47	7.856	0.860	12.51	149.9000	41.780	6.322	6.26	80.47	1.53



9.6.4 230 VAC 50 Hz, 140 V LED Load Harmonics Data

nth Order	mA Content	% Content	Limit <25 W	Remarks
1	34.90			
2	0.20	0.58%		
3	1.55	4.45%	24.8472	Pass
5	1.52	4.36%	13.8852	Pass
7	1.97	5.65%	7.3080	Pass
9	2.39	6.85%	3.6540	Pass
11	2.19	6.27%	2.5578	Pass
13	1.52	4.36%	2.1643	Pass
15	1.07	3.05%	1.8757	Pass
17	0.99	2.84%	1.6550	Pass
19	0.84	2.41%	1.4808	Pass
21	0.52	1.49%	1.3398	Pass
23	0.22	0.63%	1.2233	Pass
25	0.24	0.69%	1.1254	Pass
27	0.09	0.26%	1.0421	Pass
29	0.16	0.45%	0.9702	Pass
31	0.19	0.56%	0.9076	Pass
33	0.22	0.64%	0.8526	Pass
35	0.09	0.24%	0.8039	Pass
37	0.18	0.50%	0.7604	Pass
39	0.24	0.69%	0.7214	Pass



9.6.5 230 VAC 50 Hz, 145 V LED Load Harmonics Data

nth Order	mA Content	% Content	Limit <25 W	Remarks
1	36.09			
2	0.02	0.05%		
3	1.50	4.15%	25.8740	Pass
5	1.52	4.20%	14.4590	Pass
7	2.00	5.55%	7.6100	Pass
9	2.44	6.76%	3.8050	Pass
11	2.20	6.10%	2.6635	Pass
13	1.50	4.17%	2.2537	Pass
15	1.13	3.13%	1.9532	Pass
17	1.08	2.99%	1.7234	Pass
19	0.95	2.64%	1.5420	Pass
21	0.31	0.86%	1.3952	Pass
23	0.25	0.69%	1.2738	Pass
25	0.21	0.58%	1.1719	Pass
27	0.10	0.26%	1.0851	Pass
29	0.19	0.51%	1.0103	Pass
31	0.22	0.61%	0.9451	Pass
33	0.21	0.59%	0.8878	Pass
35	0.10	0.27%	0.8371	Pass
37	0.25	0.70%	0.7919	Pass
39	0.22	0.61%	0.7512	Pass



9.6.6 230 VAC 50 Hz, 150 V LED Load Harmonics Data

nth Order	mA Content	% Content	Limit <25 W	Remarks
1	36.73			
2	0.02	0.06%		
3	1.52	4.14%	26.4826	Pass
5	1.54	4.18%	14.7991	Pass
7	2.03	5.54%	7.7890	Pass
9	2.48	6.74%	3.8945	Pass
11	2.25	6.13%	2.7262	Pass
13	1.54	4.18%	2.3067	Pass
15	1.12	3.04%	1.9992	Pass
17	1.09	2.97%	1.7640	Pass
19	0.91	2.49%	1.5783	Pass
21	0.36	0.97%	1.4280	Pass
23	0.25	0.69%	1.3038	Pass
25	0.22	0.60%	1.1995	Pass
27	0.06	0.16%	1.1107	Pass
29	0.18	0.49%	1.0341	Pass
31	0.27	0.72%	0.9673	Pass
33	0.23	0.62%	0.9087	Pass
35	0.08	0.21%	0.8568	Pass
37	0.25	0.68%	0.8105	Pass
39	0.23	0.62%	0.7689	Pass



10 Thermal Performance

Images captured after running for more than 30 minutes at room temperature (25 °C), open frame.

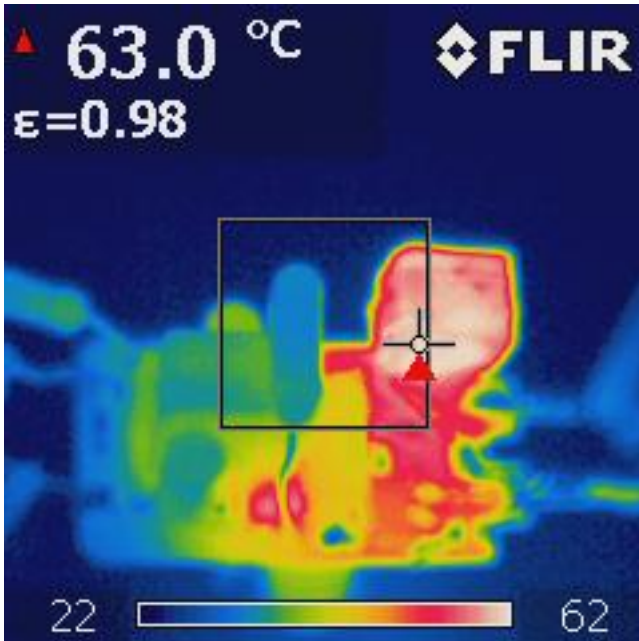


Figure 16 – 195 VAC, Full Load.
Transformer Temperature = 63 °C.

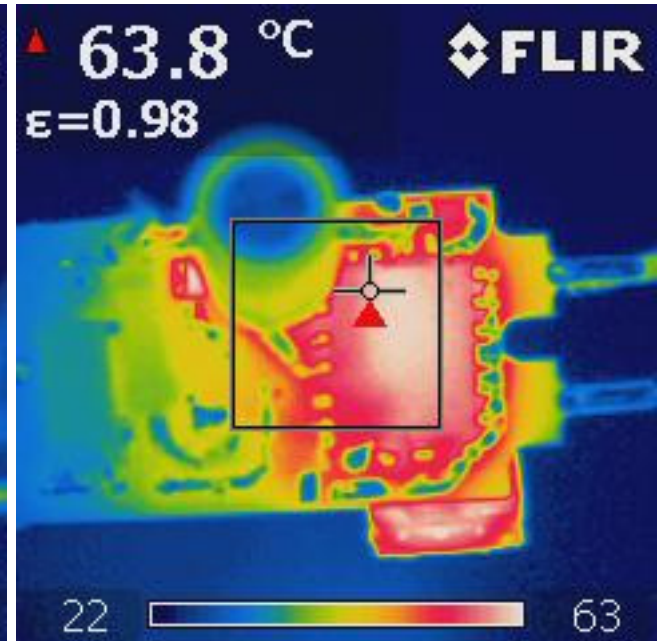


Figure 17 – 195 VAC, Full Load.
LNK460KG Temperature = 63 °C.

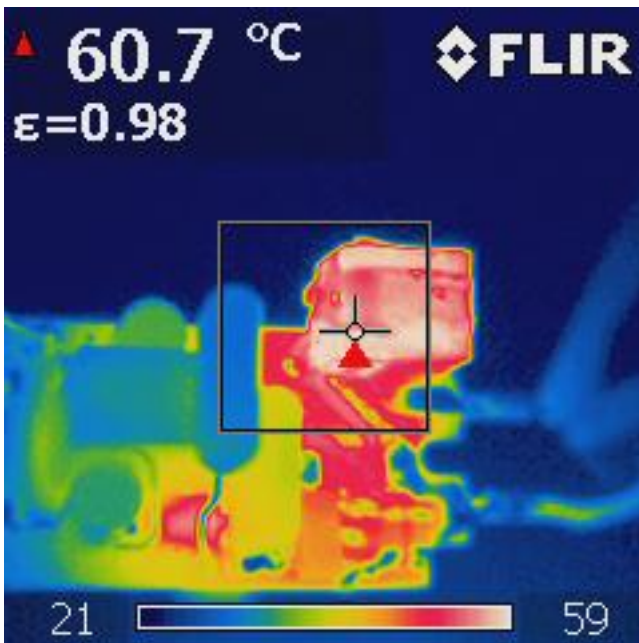


Figure 18 – 230 VAC, Full Load.
Transformer Temperature = 61 °C.

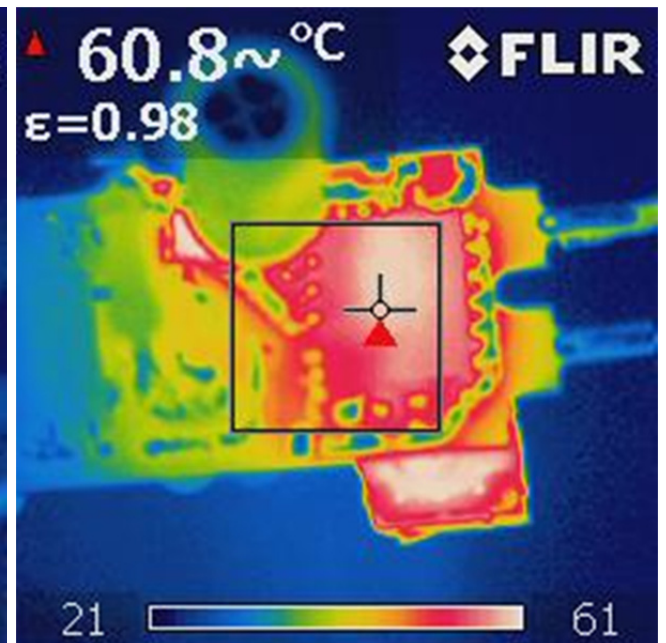


Figure 19 – 230 VAC, Full Load.
LNK460KG Temperature = 61 °C.

11 Waveforms

11.1 Input Voltage and Input Current Waveforms

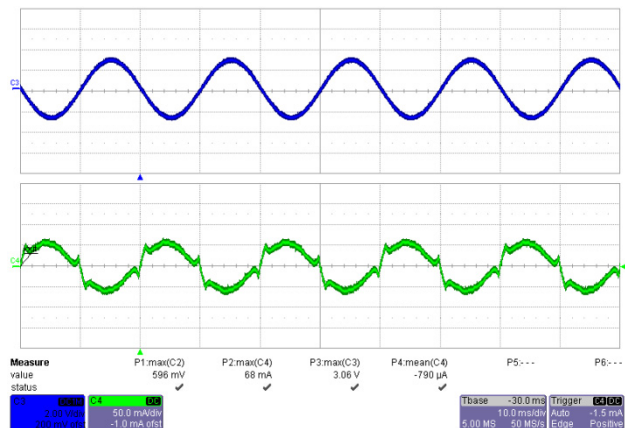


Figure 20 – 195 VAC, 50 Hz Full Load.
Upper: V_{IN} , 200 V / div.
Lower: I_{IN} , 50 mA, 10 ms / div.

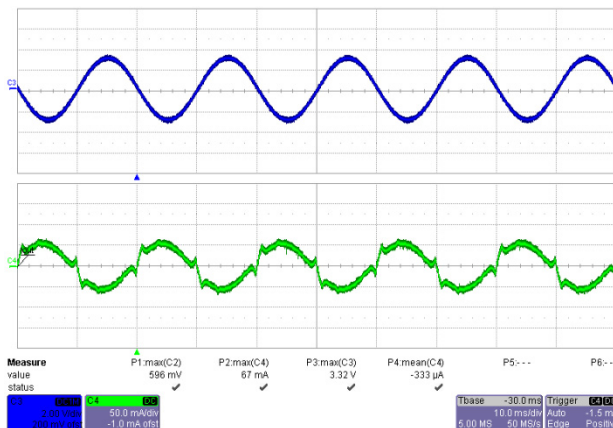


Figure 21 – 210 VAC, 50 Hz Full Load.
Upper: V_{IN} , 200 V / div.
Lower: I_{IN} , 50 mA, 10 ms / div.

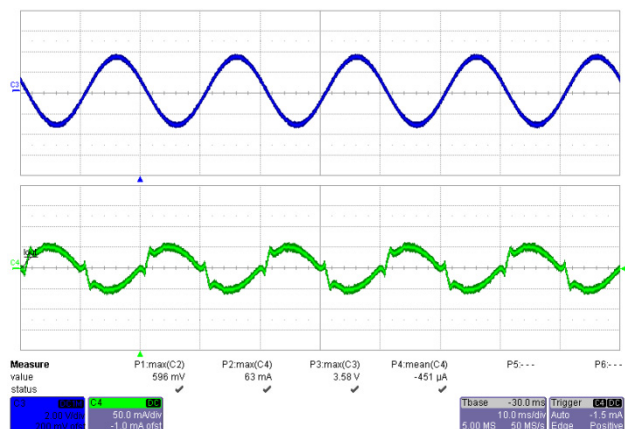


Figure 22 – 230 VAC, 50 Hz Full Load.
Upper: V_{IN} , 200 V / div.
Lower: I_{IN} , 50 mA, 10 ms / div.

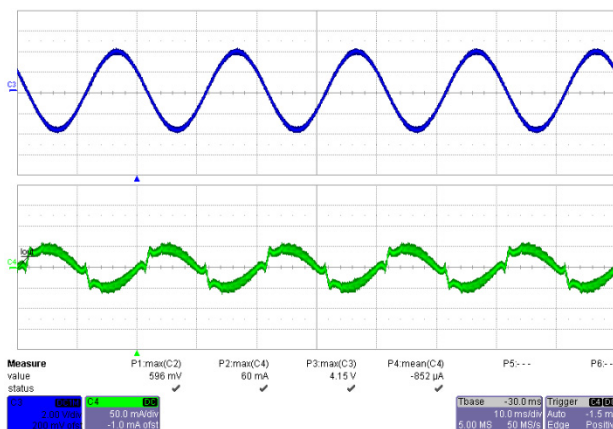


Figure 23 – 265 VAC, 50 Hz Full Load.
Upper: V_{IN} , 200 V / div.
Lower: I_{IN} , 50 mA, 10 ms / div.



11.2 Output Current and Output Voltage at Normal Operation

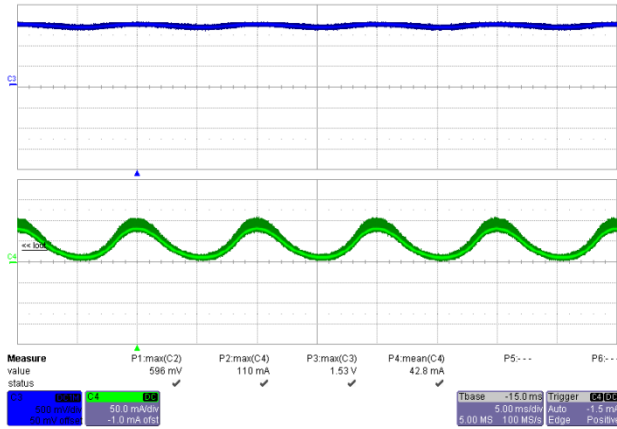


Figure 24 – 195 VAC, 50 Hz Full Load.
Upper: V_{OUT} , 50 V / div.
Lower: I_{OUT} , 50 mA, 5 ms / div.

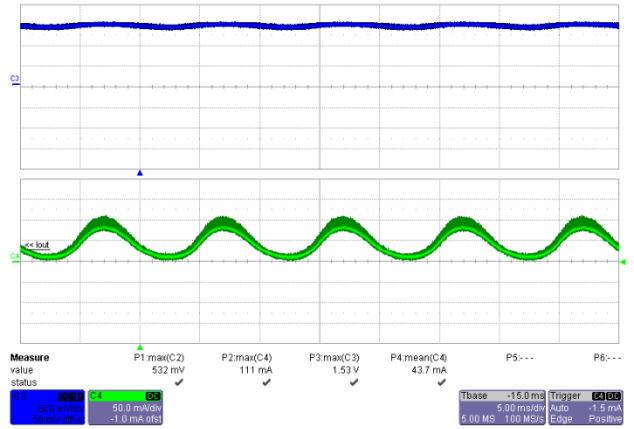


Figure 25 – 210 VAC, 50 Hz Full Load.
Upper: V_{OUT} , 50 V / div.
Lower: I_{OUT} , 50 mA, 5 ms / div.

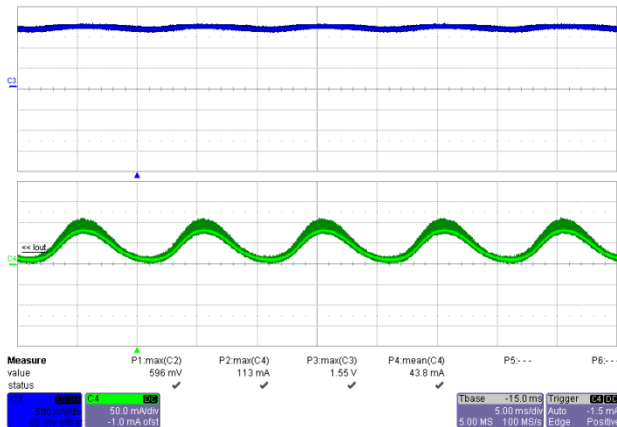


Figure 26 – 230 VAC, 50 Hz Full Load.
Upper: V_{OUT} , 50 V / div.
Lower: I_{OUT} , 50 mA, 5 ms / div.

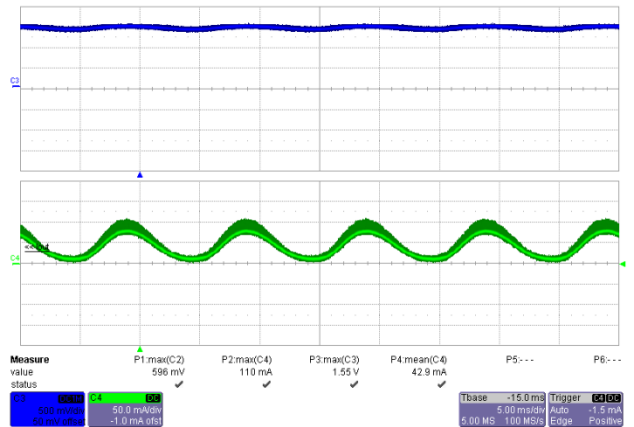


Figure 27 – 265 VAC, 50 Hz Full Load.
Upper: V_{OUT} , 50 V / div.
Lower: I_{OUT} , 50 mA, 5 ms / div.



11.3 Output Current/Voltage Rise and Fall

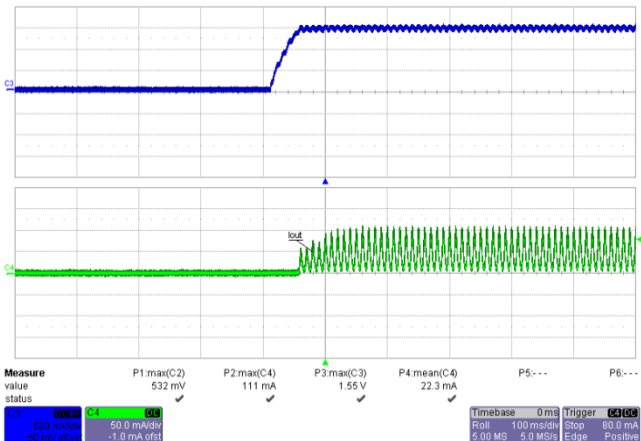


Figure 28 – 195 VAC, 50 Hz Output Rise.
Upper: V_{OUT} , 50 V / div.
Lower: I_{OUT} , 50 mA, 100 ms / div.

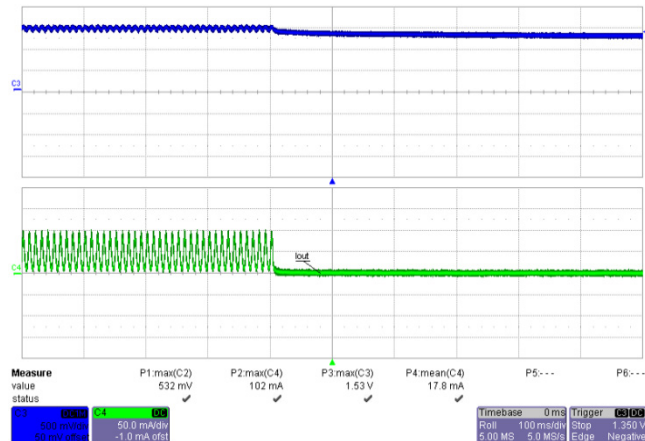


Figure 29 – 195 VAC, 50 Hz Output Fall.
Upper: V_{OUT} , 50 V / div.
Lower: I_{OUT} , 50 mA, 100 ms / div.

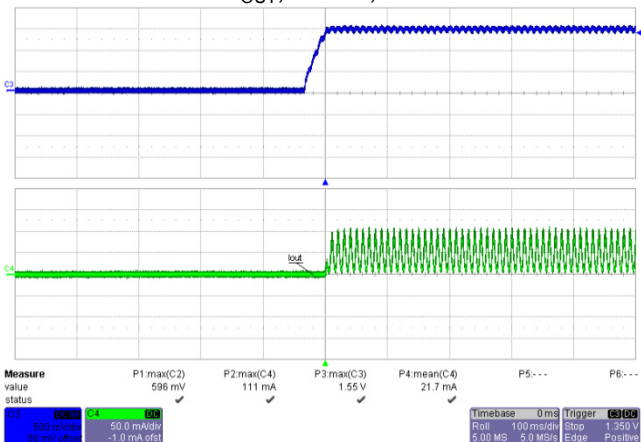


Figure 30 – 265 VAC, 50 Hz Output Rise.
Upper: V_{OUT} , 50 V / div.
Lower: I_{OUT} , 50 mA, 100 ms / div.

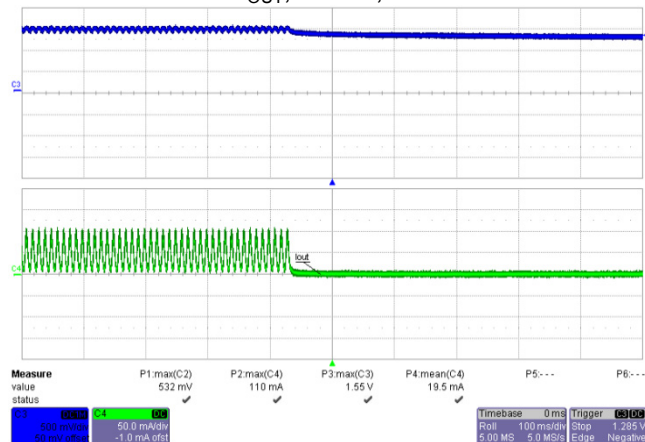


Figure 31 – 265 VAC, 50 Hz Output Fall.
Upper: V_{OUT} , 50 V / div.
Lower: I_{OUT} , 50 mA, 100 ms / div.



11.4 Input Voltage and Output Current Waveform at Start-up

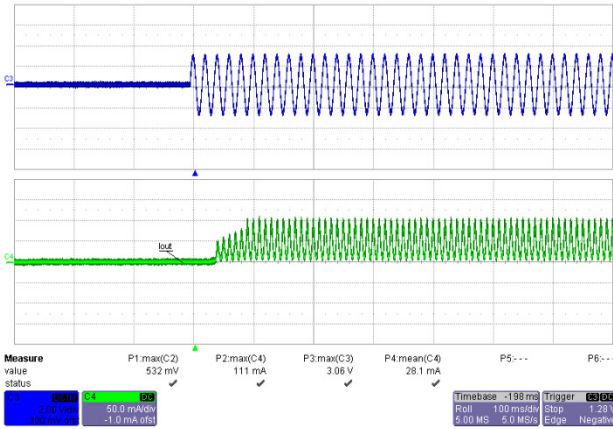


Figure 32 – 195 VAC, 50 Hz.
 Upper: V_{IN} , 200 V / div.
 Lower: I_{OUT} , 50 mA, 100 ms / div.

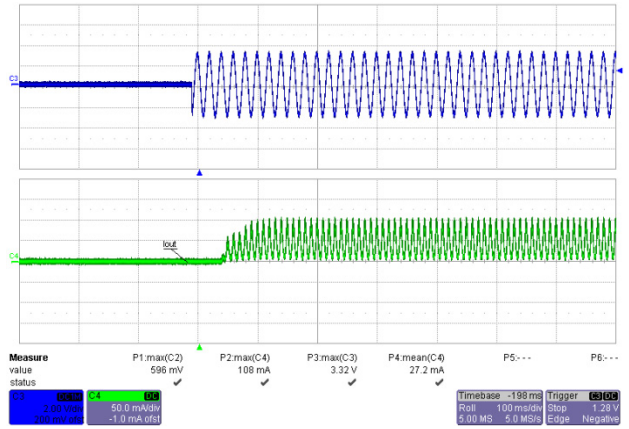


Figure 33 – 210 VAC, 50 Hz.
 Upper: V_{IN} , 200 V / div.
 Lower: I_{OUT} , 50 mA, 100 ms / div.

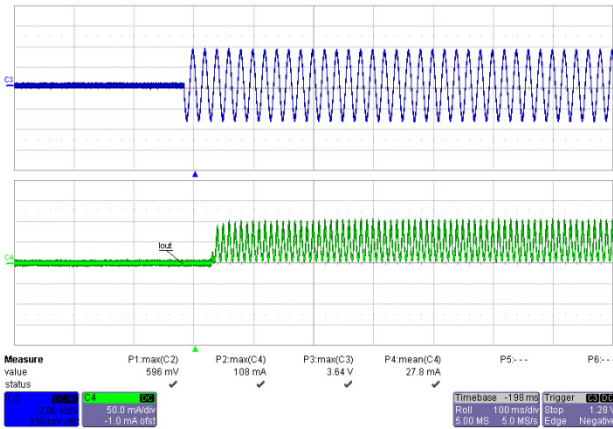


Figure 34 – 230 VAC, 50 Hz.
 Upper: V_{IN} , 200 V / div.
 Lower: I_{OUT} , 50 mA, 100 ms / div.

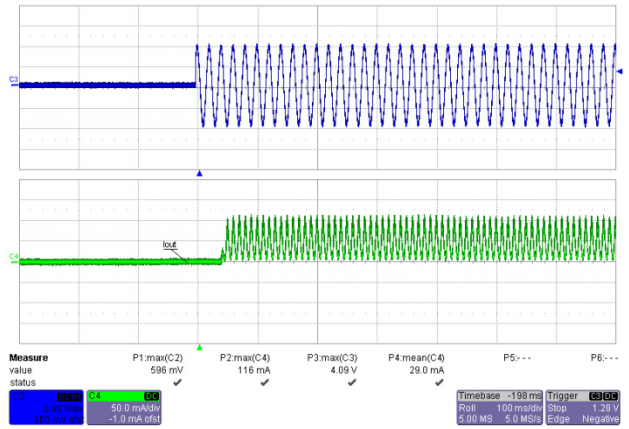


Figure 35 – 265 VAC, 50 Hz.
 Upper: V_{IN} , 200 V / div.
 Lower: I_{OUT} , 50 mA, 100 ms / div.

11.5 Drain Voltage and Current at Normal Operation

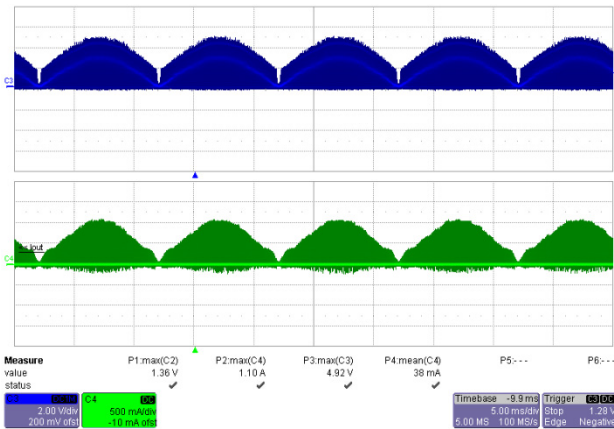


Figure 36 – 195 VAC, 50 Hz.
Upper: V_{DRAIN} , 200 V / div.
Lower: I_{DRAIN} , 500 mA, 5 ms / div.

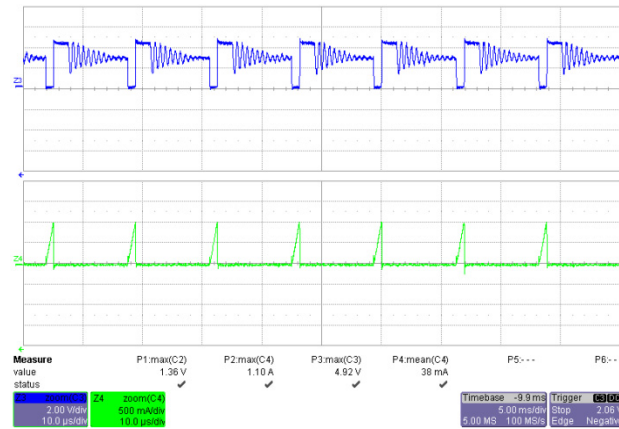


Figure 37 – 195 VAC, 50 Hz.
Upper: V_{DRAIN} , 200 V / div.
Lower: I_{DRAIN} , 500 mA, 5 μ s / div.

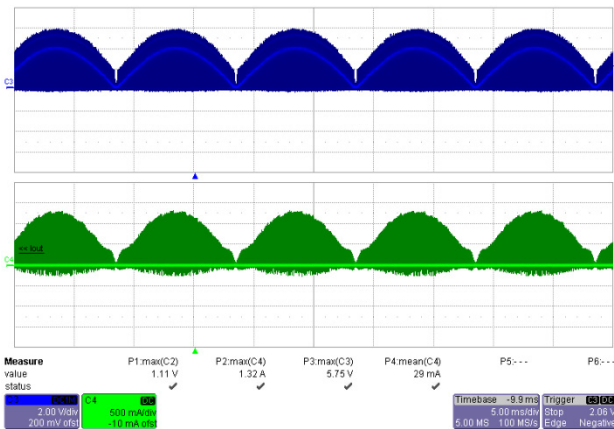


Figure 38 – 265 VAC, 50 Hz.
Upper: V_{DRAIN} , 200 V / div.
Lower: I_{DRAIN} , 500 mA, 5 ms / div.

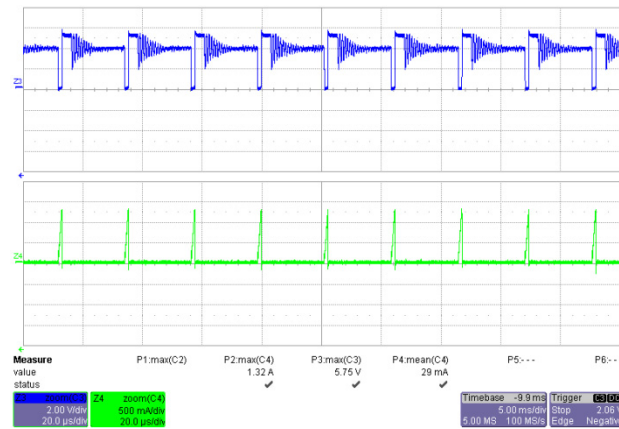


Figure 39 – 265 VAC, 50 Hz.
Upper: V_{DRAIN} , 200 V / div.
Lower: V_{DRAIN} , 500 mA, 5 μ s / div.

11.6 Start-up Drain Voltage and Current

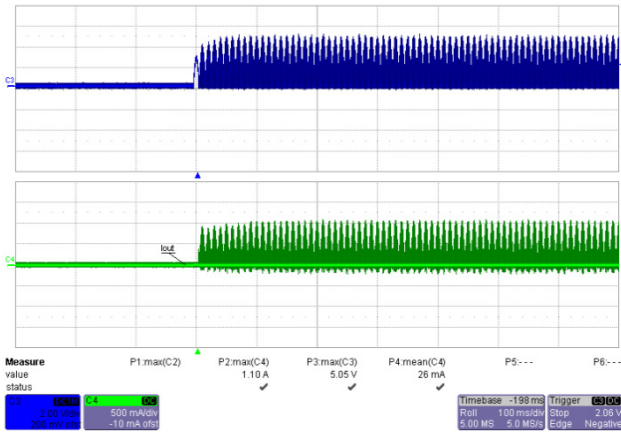


Figure 40 – 195 VAC, 50 Hz.
 Upper: V_{DRAIN} , 200 V / div.
 Lower: I_{DRAIN} , 500 mA, 100 ms / div.

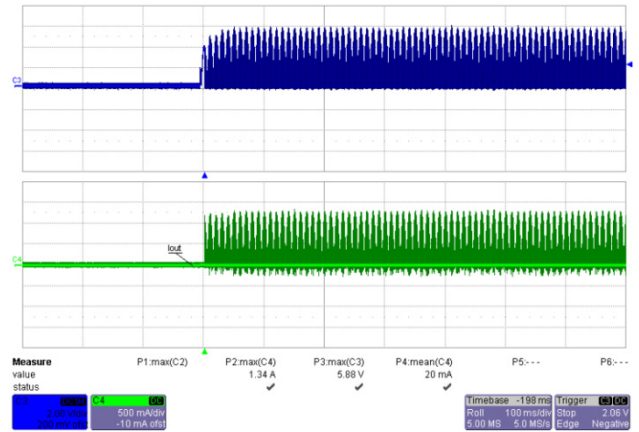


Figure 41 – 265 VAC, 50 Hz.
 Upper: V_{DRAIN} , 200 V / div.
 Lower: I_{DRAIN} , 500 mA, 100 ms / div.

11.7 Drain Current and Drain Voltage During Output Short Condition

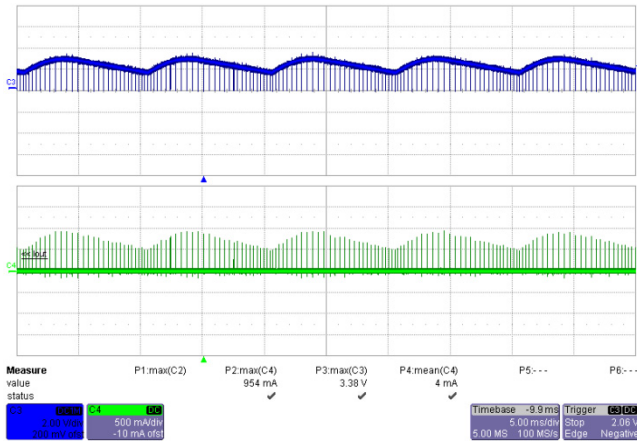


Figure 42 – 195 VAC, 50 Hz Output Short Condition.
 Upper: V_{DRAIN} , 200 V / div.
 Lower: I_{DRAIN} , 500 mA, 5 ms / div.

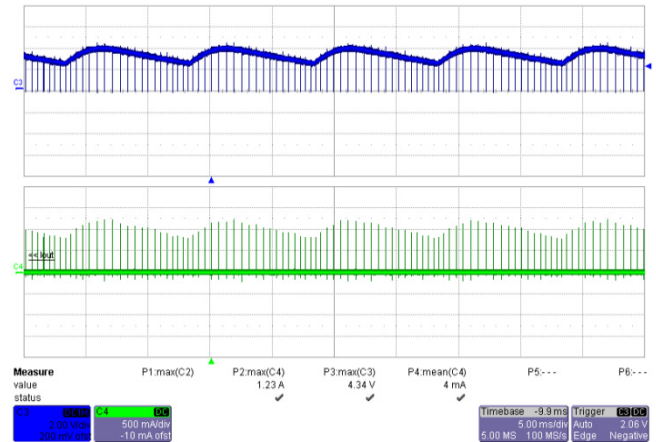


Figure 43 – 265 VAC, 50 Hz Output Short Condition.
 Upper: V_{DRAIN} , 200 V / div.
 Lower: I_{DRAIN} , 500 mA, 5 ms / div.



11.8 Freewheeling Diode Voltage and Output Current Waveform

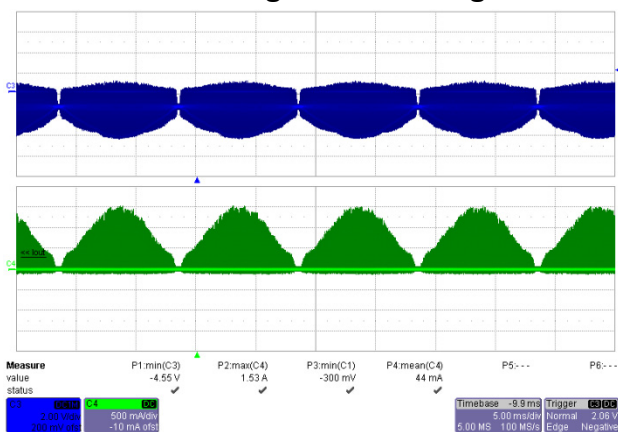


Figure 44 – 195 VAC, 50 Hz.
 Upper: V_{IN} , 200 V / div.
 Lower: I_{OUT} , 50 mA, 100 ms / div.

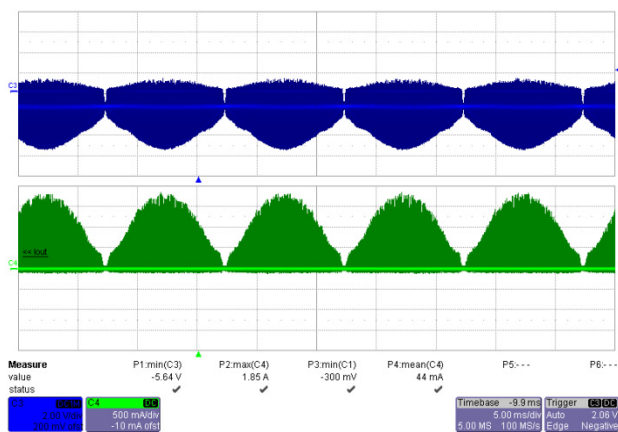


Figure 45 – 210 VAC, 50 Hz.
 Upper: V_{IN} , 200 V / div.
 Lower: I_{OUT} , 50 mA, 100 ms / div.

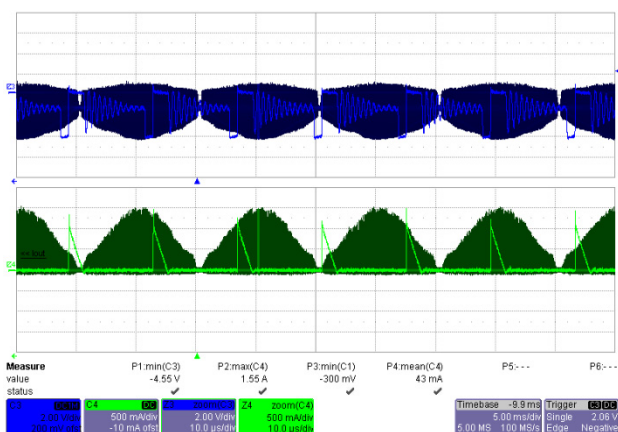


Figure 46 – 230 VAC, 50 Hz.
 Upper: V_{IN} , 200 V / div.
 Lower: I_{OUT} , 50 mA, 100 ms / div.

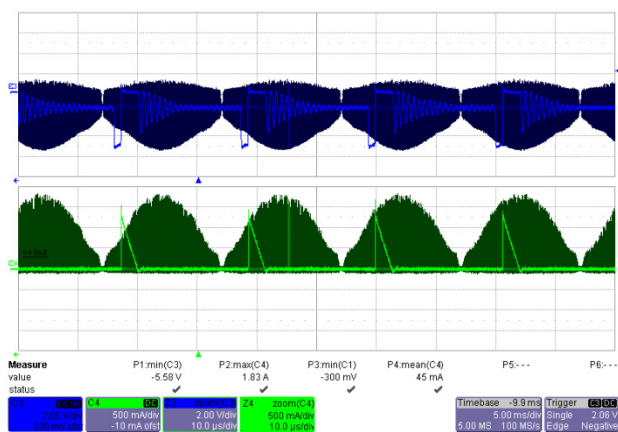


Figure 47 – 265 VAC, 50 Hz.
 Upper: V_{IN} , 200 V / div.
 Lower: I_{OUT} , 50 mA, 100 ms / div.

11.9 Input Voltage and Output Current Waveform with a Leading Edge Dimmer

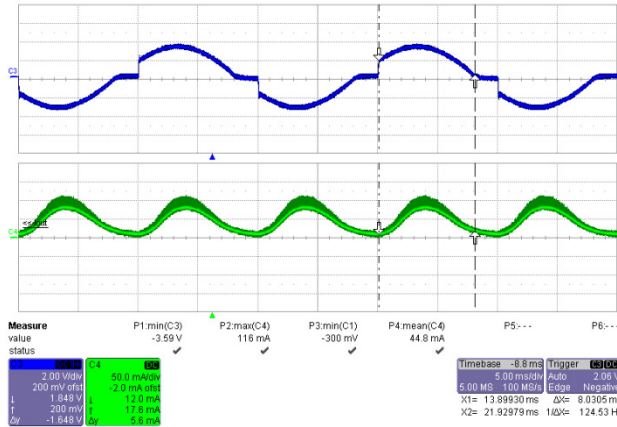


Figure 48 – 230 VAC, 50 Hz.
Upper: V_{IN} , 200 V / div.
Lower: I_{OUT} , 50 mA, 100 ms / div.

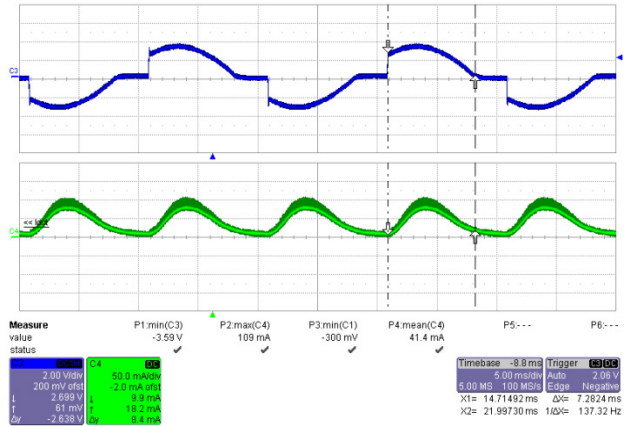


Figure 49 – 230 VAC, 50 Hz.
Upper: V_{IN} , 200 V / div.
Lower: I_{OUT} , 50 mA, 100 ms / div.

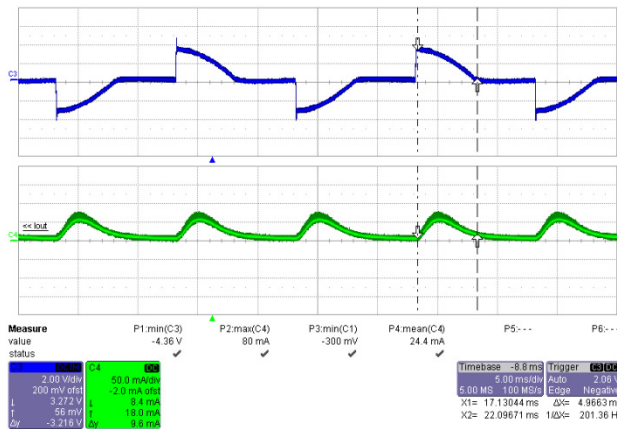


Figure 50 – 230 VAC, 50 Hz.
Upper: V_{IN} , 200 V / div.
Lower: I_{OUT} , 50 mA, 100 ms / div.

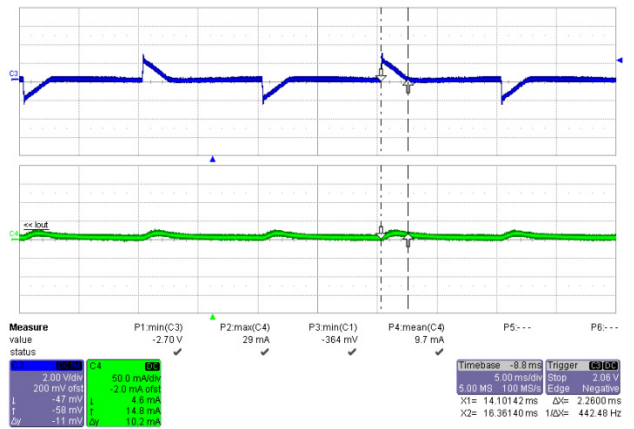


Figure 51 – 230 VAC, 50 Hz.
Upper: V_{IN} , 200 V / div.
Lower: I_{OUT} , 50 mA, 100 ms / div.



11.10 Input Voltage and Output Current Waveform with a Trailing Edge Dimmer

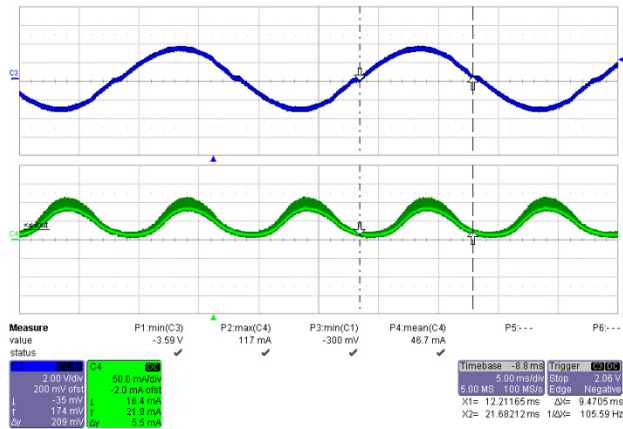


Figure 52 – 230 VAC, 50 Hz.
Upper: V_{IN} , 200 V / div.
Lower: I_{OUT} , 50 mA, 100 ms / div.

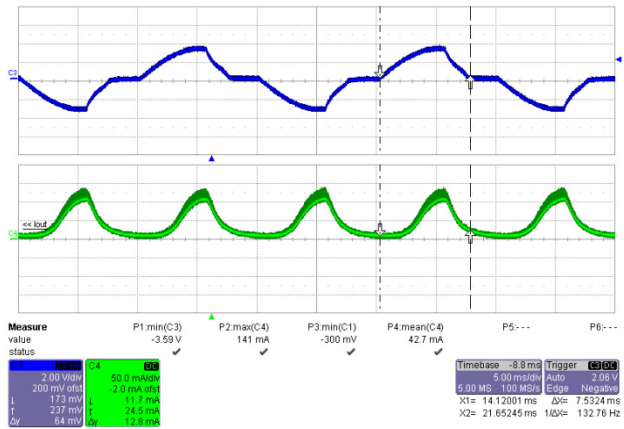


Figure 53 – 230 VAC, 50 Hz.
Upper: V_{IN} , 200 V / div.
Lower: I_{OUT} , 50 mA, 100 ms / div.

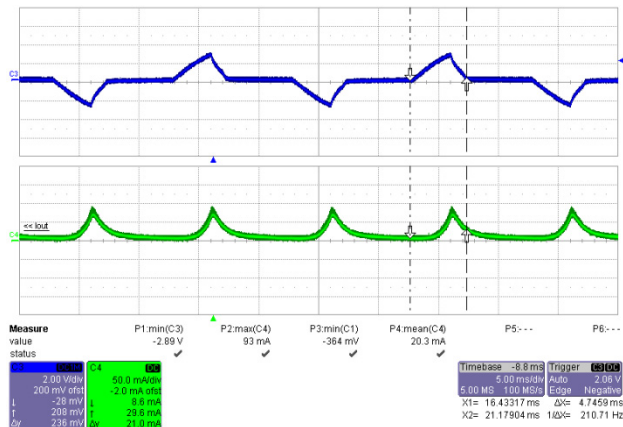


Figure 54 – 230 VAC, 50 Hz.
Upper: V_{IN} , 200 V / div.
Lower: I_{OUT} , 50 mA, 100 ms / div.



11.11 Voltage Waveforms Across Damping Resistor and Bleeder Resistor

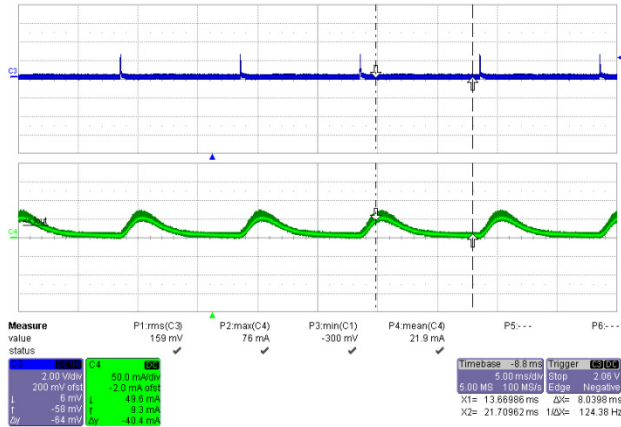


Figure 55 – 230 VAC, 50 Hz. at 90° Phase.
 Upper: V_{IN} , 200 V / div.
 Lower: I_{OUT} , 50 mA, 100 ms / div.

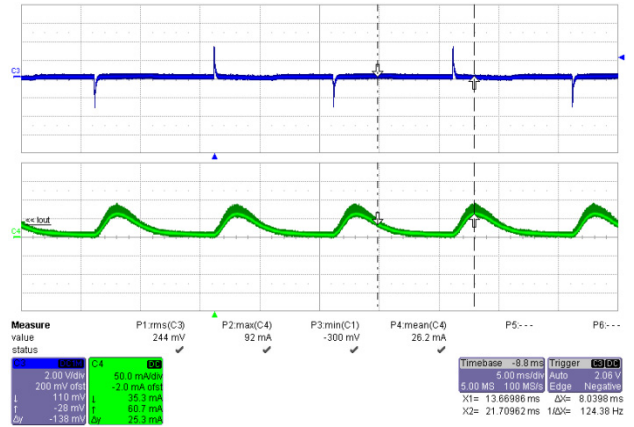


Figure 56 – 230 VAC, 50 Hz. at 90° Phase.
 Upper: V_{IN} , 200 V / div.
 Lower: I_{OUT} , 50 mA, 100 ms / div.

11.12 Output Voltage Waveform in Open Load Condition

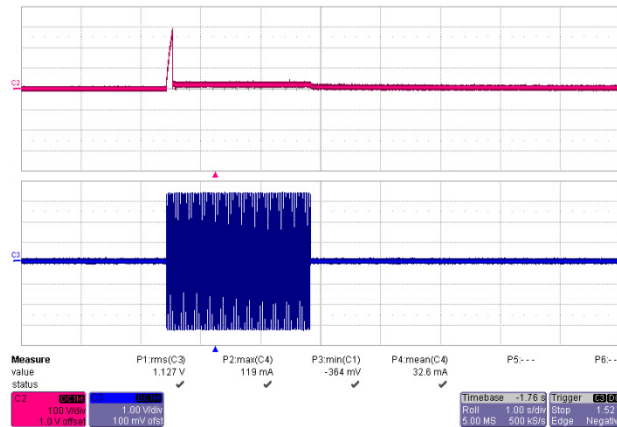
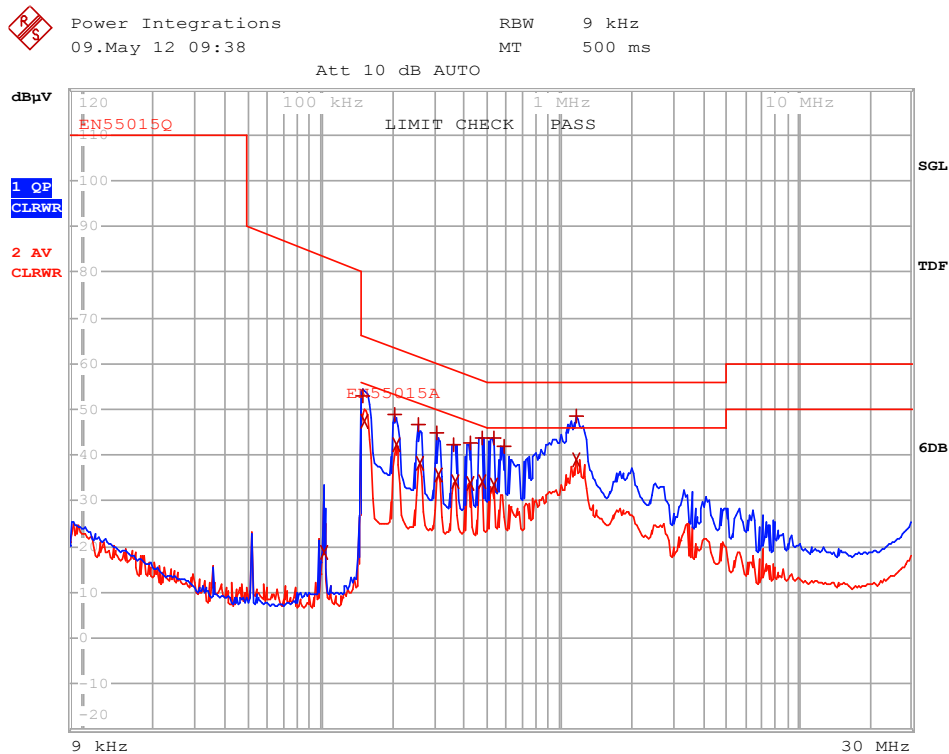


Figure 57 – 230 VAC, 50 Hz.
 Upper: V_{OUT} , 100 V / div.
 Lower: V_{IN} , 100 V, 1 s / div.



12 Conducted EMI



EDIT PEAK LIST (Final Measurement Results)

```
Trace1: EN55015Q
Trace2: EN55015A
Trace3: ---
```

TRACE	FREQUENCY	LEVEL dBµV	DELTA	LIMIT dB
2 Average	104.063986756 kHz	18.68	L1 gnd	
1 Quasi Peak	151.5 kHz	53.02	L1 gnd	-12.89
2 Average	153.015 kHz	47.48	L1 gnd	-8.35
1 Quasi Peak	204.199110673 kHz	48.82	L1 gnd	-14.61
2 Average	208.303512797 kHz	42.37	N gnd	-10.89
1 Quasi Peak	256.711570318 kHz	46.55	N gnd	-14.98
2 Average	259.278686021 kHz	38.14	N gnd	-13.31
1 Quasi Peak	307.064896815 kHz	44.70	N gnd	-15.34
2 Average	310.135545783 kHz	35.71	N gnd	-14.25
1 Quasi Peak	360.057740611 kHz	42.33	N gnd	-16.39
2 Average	367.294901197 kHz	34.06	N gnd	-14.50
2 Average	422.19601758 kHz	33.88	N gnd	-13.52
1 Quasi Peak	426.417977756 kHz	42.72	N gnd	-14.60
1 Quasi Peak	475.741040231 kHz	43.81	N gnd	-12.59
2 Average	475.741040231 kHz	34.02	N gnd	-12.38
1 Quasi Peak	530.769219795 kHz	43.71	N gnd	-12.28
2 Average	530.769219795 kHz	33.30	N gnd	-12.69
1 Quasi Peak	586.299423673 kHz	41.93	N gnd	-14.06
1 Quasi Peak	1.17656420634 MHz	48.39	N gnd	-7.61
2 Average	1.17656420634 MHz	39.06	N gnd	-6.93

Figure 58 – Conducted EMI, 145 V LED Load, 230 VAC, 50 Hz, and EN55015 B Limits.



13 Revision History

Date	Author	Revision	Description and Changes	Reviewed
10-Jun-13	DK	1.0	Initial Release	Apps & Mktg



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